

Electrochemistry for Materials Technology

Chapter Z: Electrochemical Impedance Spectroscopy (EIS)

Part I: Fundamentals of Dielectric Spectroscopy

- 1 – Electrochemical interfacial processes
- 2 – Principle of impedance spectroscopy
- 3 – Ideal Circuit elements
- 4 – Case study 1: the RC element
- 5 – Representing EIS data: Bode and Nyquist plots
- 6 – Treatment of a ZARC element
- 7 – Faradaic reaction: the Randles circuit
- 8 – Non ideal elements

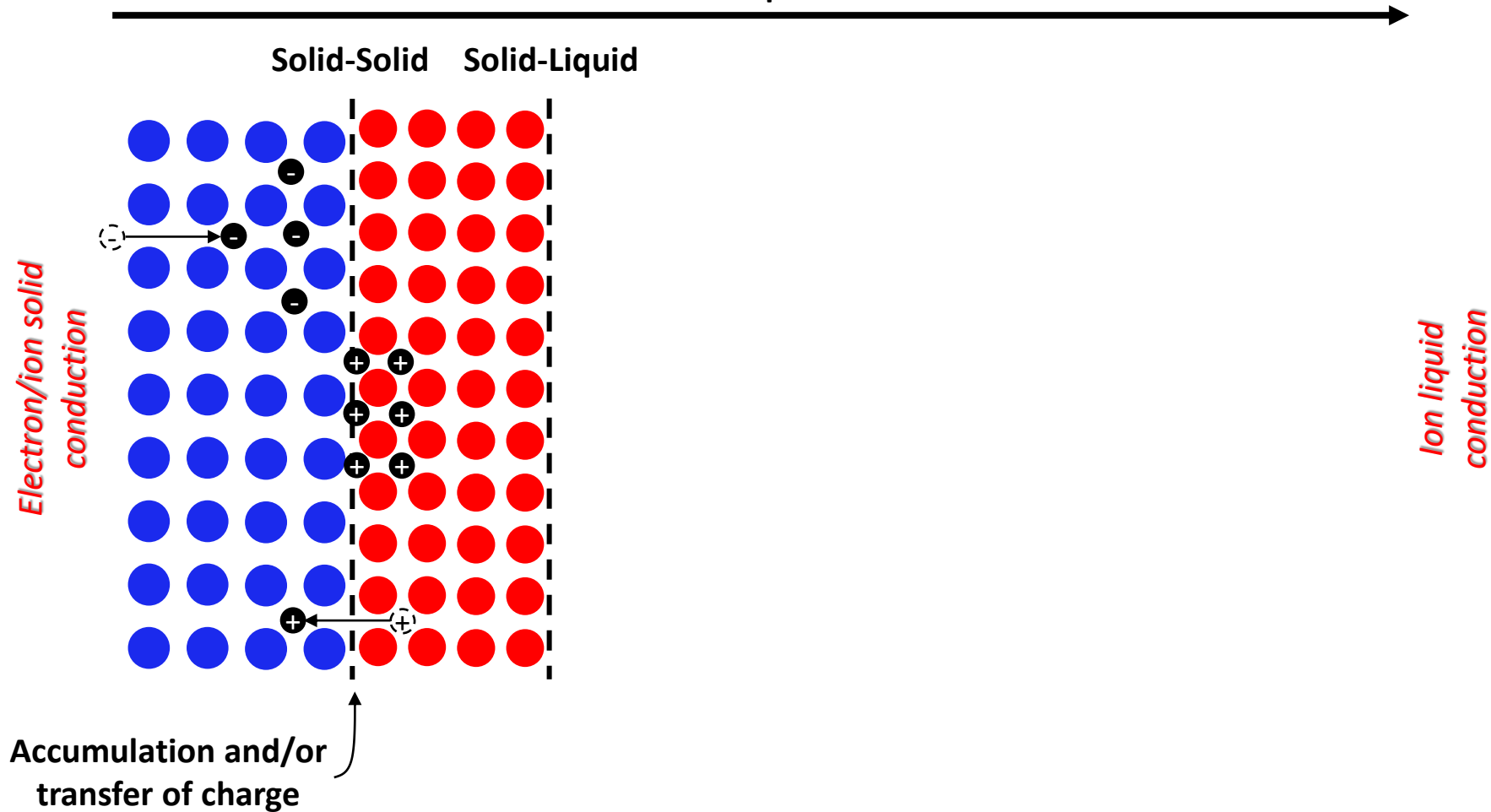
Part II: Practical aspects and data treatment

- 1 – Impedance measurements
- 2 – Data treatment
- 3 – Non-linear systems: Total Harmonic Distorsion

I) Fundamentals of EIS

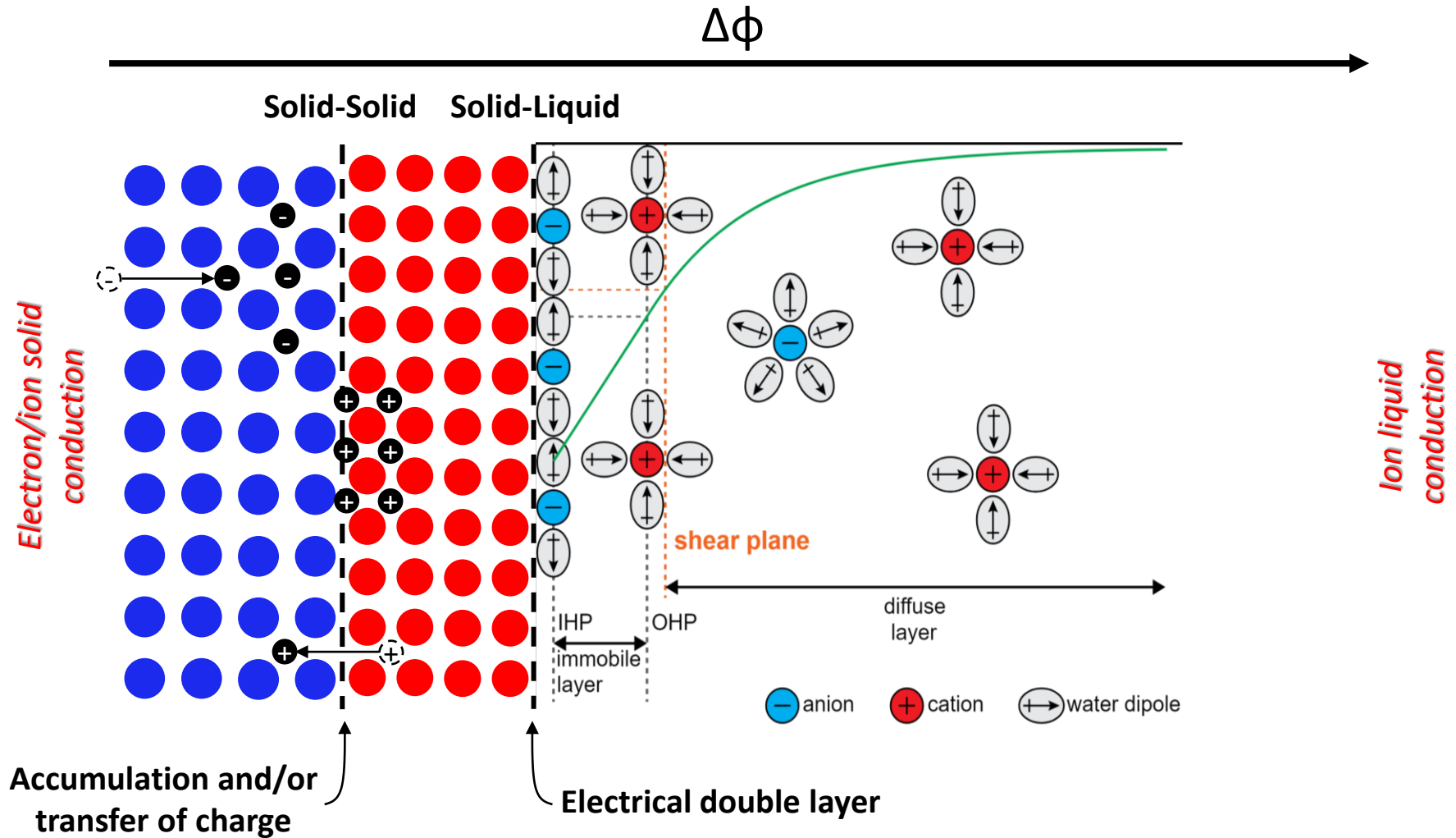
1) Electrochemical interfacial processes

$\Delta\phi$



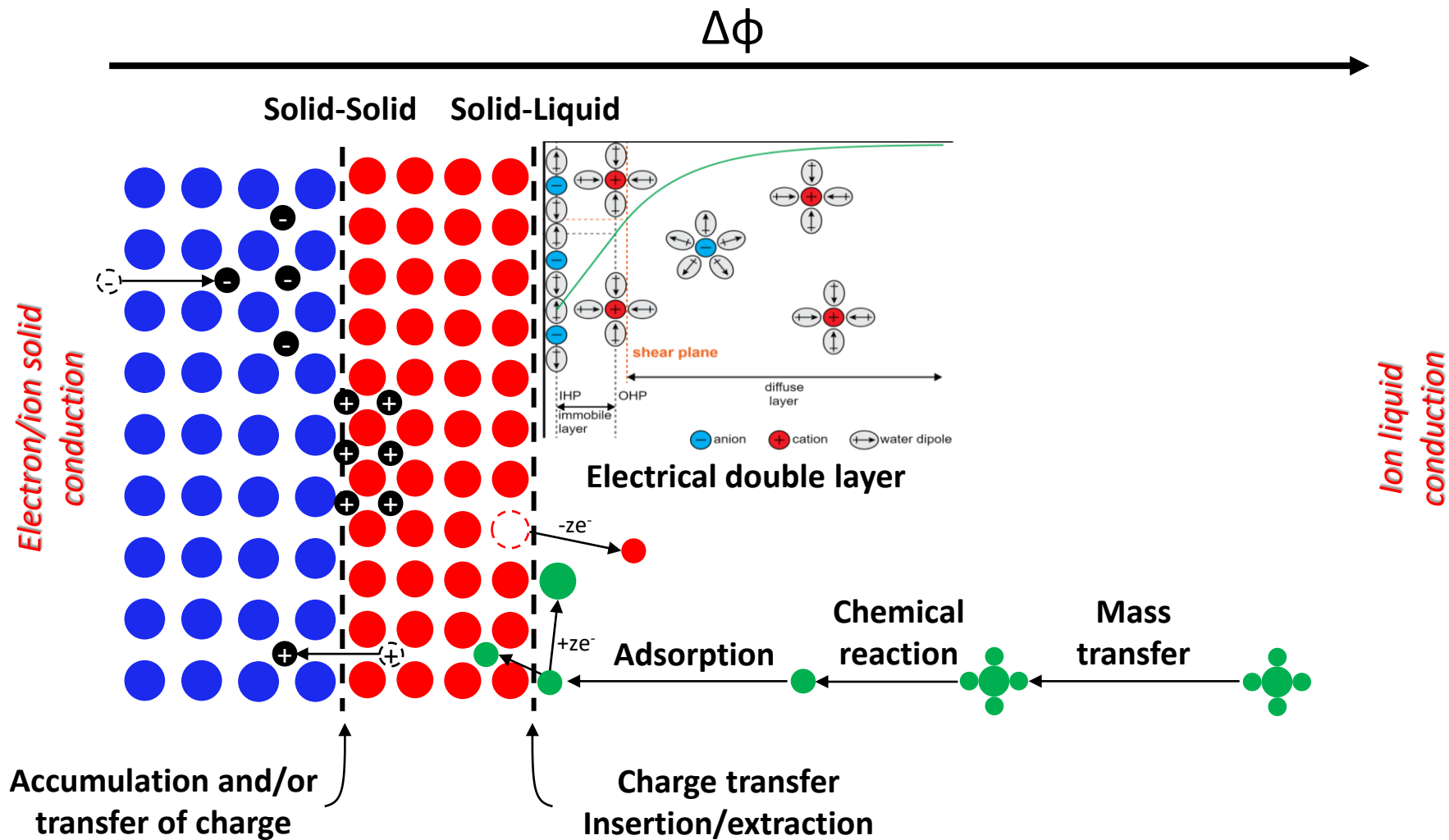
I) Fundamentals of EIS

1) Electrochemical interfacial processes



I) Fundamentals of EIS

1) Electrochemical interfacial processes

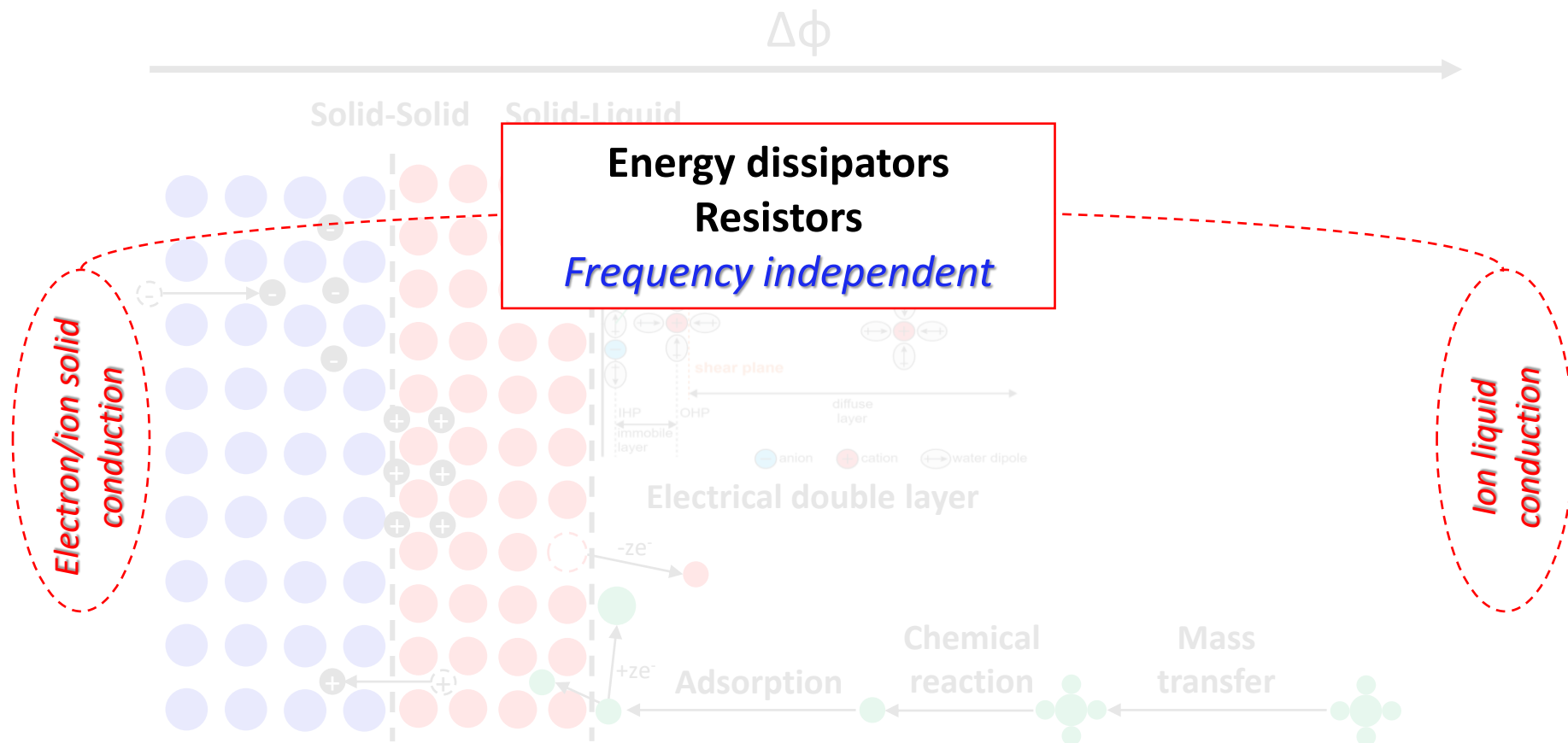


Process rates

$\approx 1 \text{ MHz} - 100 \text{ kHz} - 1 \text{ kHz} - 10 \text{ Hz} - 100 \text{ mHz}$

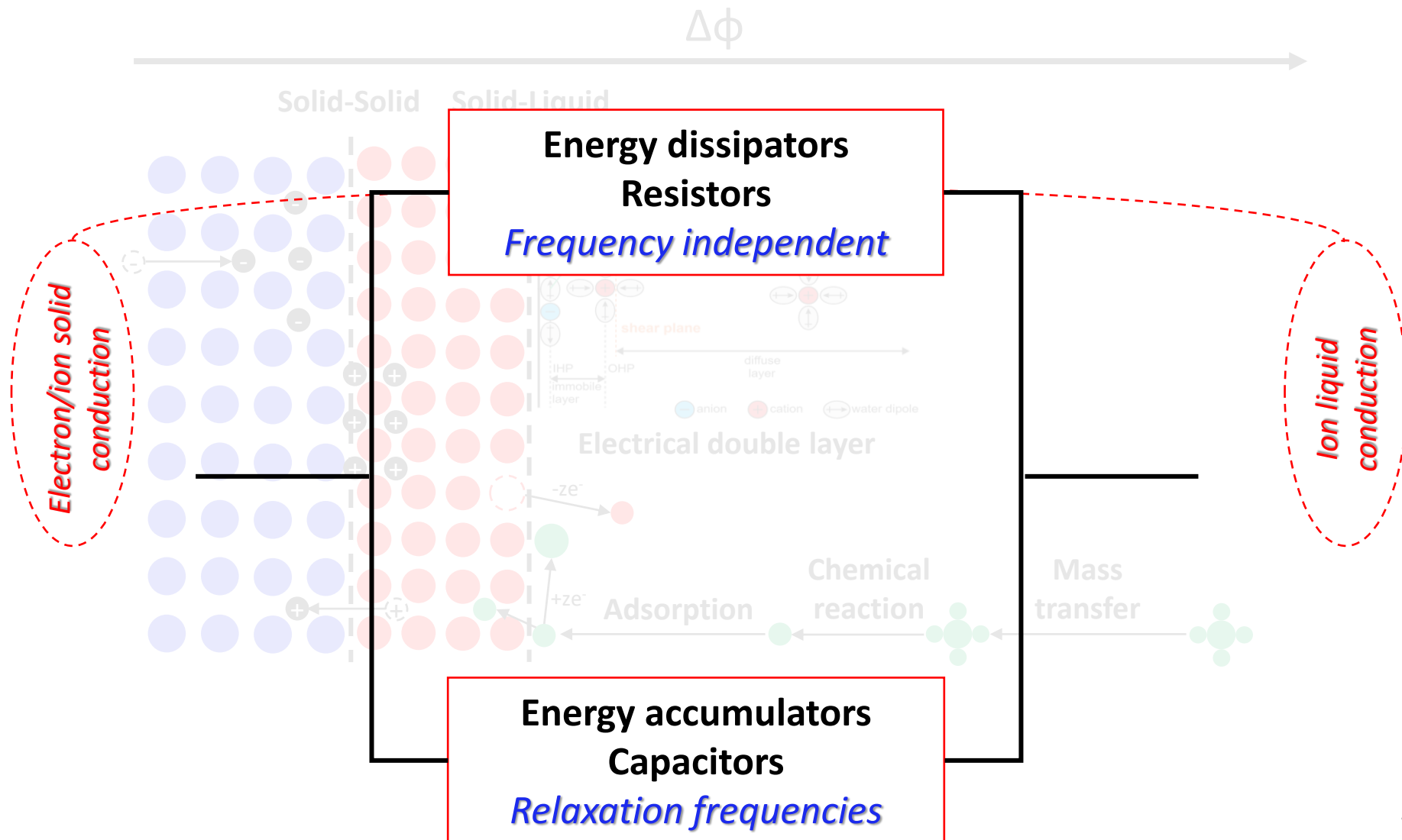
I) Fundamentals of EIS

1) Electrochemical interfacial processes



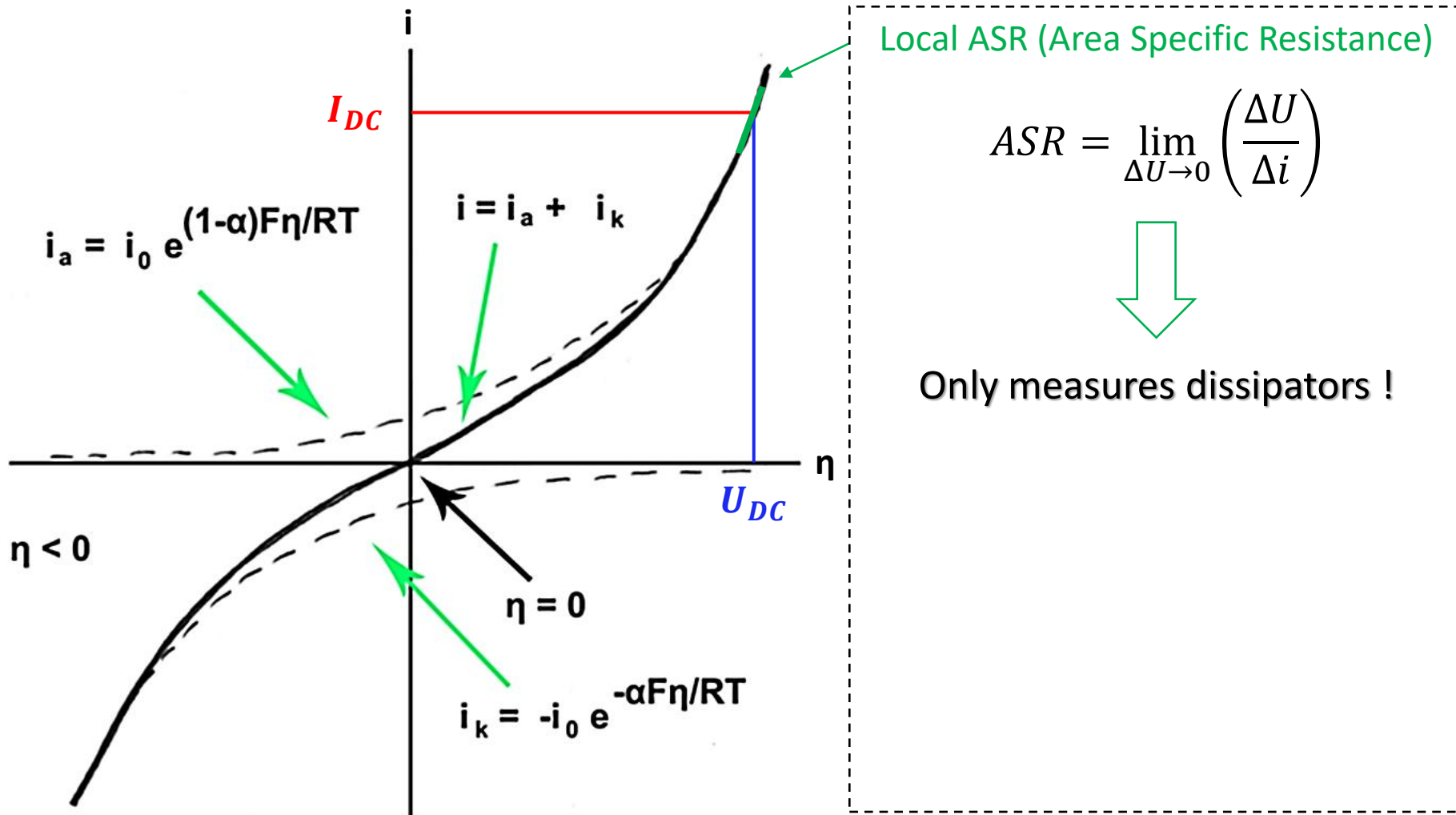
I) Fundamentals of EIS

1) Electrochemical interfacial processes



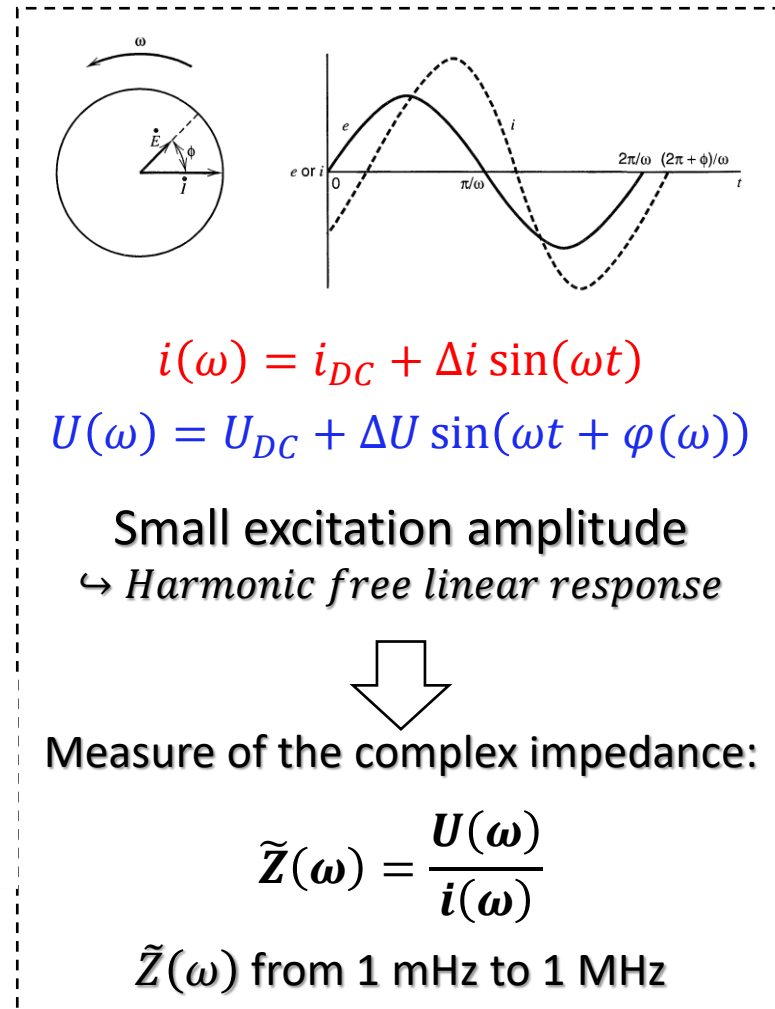
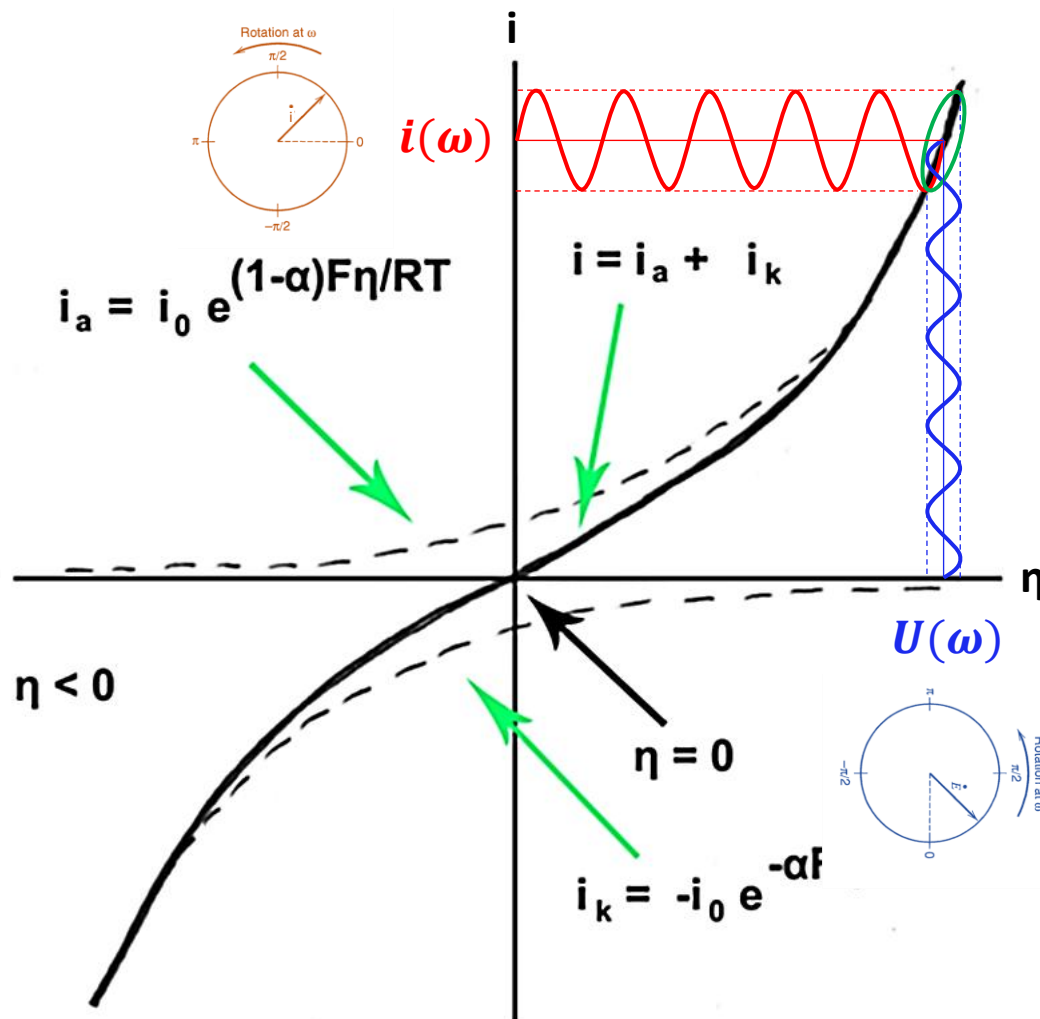
1) Fundamentals of EIS

2) Principle of impedance spectroscopy



1) Fundamentals of EIS

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1) Fundamentals of EIS

2) Principle of impedance spectroscopy

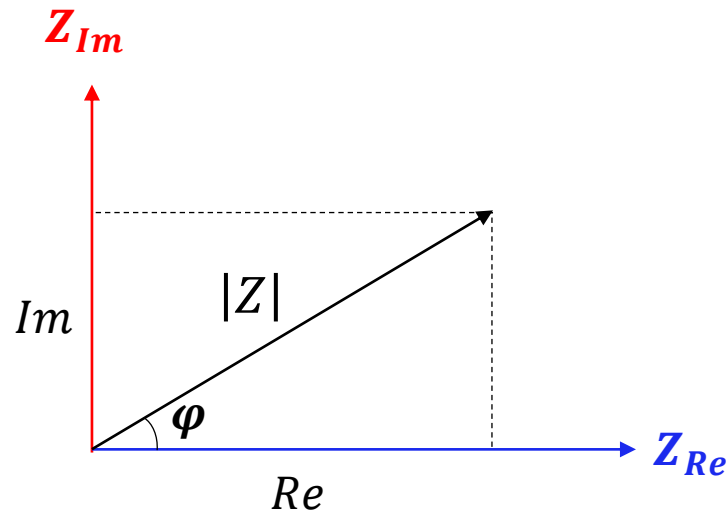
$$\tilde{Z}(\omega) = \frac{U_0 \sin(\omega t + \varphi)}{i_0 \sin(\omega t)}$$

$$\tilde{Z}(\omega) = \frac{U_0}{i_0} \cdot \frac{e^{j(\omega t + \varphi)} - e^{-j(\omega t + \varphi)}}{e^{j\omega t} - e^{-j\omega t}} = \frac{U_0}{i_0} \cdot e^{j\varphi}$$

$$\tilde{Z}(\omega) = \frac{U_0}{i_0} \cdot (\cos \varphi + j \sin \varphi)$$


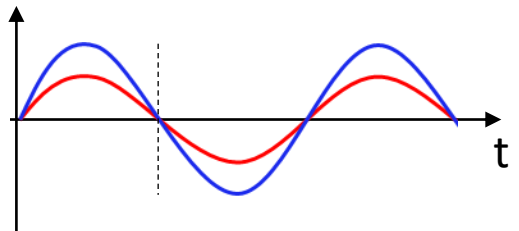

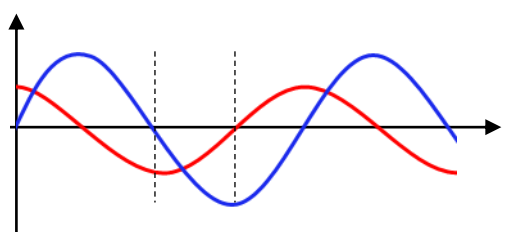

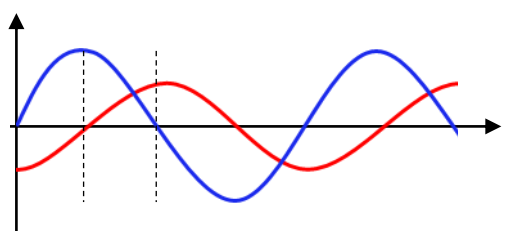
$$\tilde{Z}(\omega) = Z_{Re} + jZ_{Im}$$

Vector representation:



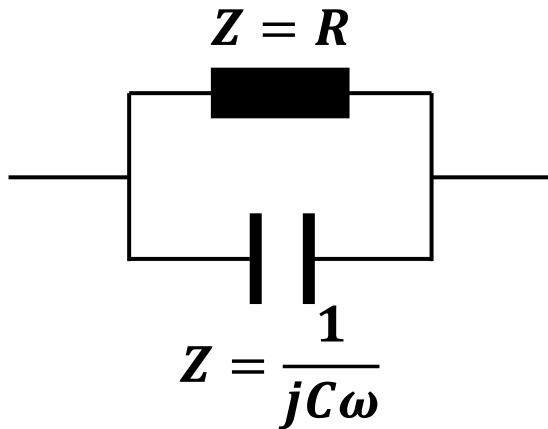
1) Fundamentals of EIS

3) Ideal circuit elements

Element	$U(\omega), i(\omega)$	Phase shift φ	$\tilde{Z}(\omega)$
Resistor 		$\varphi = 0$ Current and voltage in phase	$\tilde{Z}(\omega) = R$
Capacitor 		$\varphi = -\frac{\pi}{2}$ Current leads the voltage	$\tilde{Z}(\omega) = -\frac{j}{C\omega}$
Inductance 		$\varphi = +\frac{\pi}{2}$ Current lags behind the voltage	$\tilde{Z}(\omega) = jL\omega$

I) Fundamentals of EIS

4) The RC element



$$\frac{1}{Z} = \frac{1}{R} + jC\omega = \frac{1 + jRC\omega}{R}$$

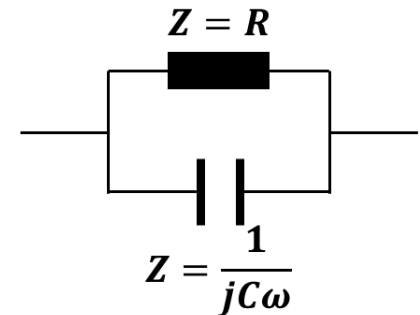
$$Z = \frac{R}{1 + jRC\omega} = \frac{R - jR^2C\omega}{1 + (RC\omega)^2}$$

$$Z = \frac{R}{1 + (RC\omega)^2} - j \frac{R^2C\omega}{1 + (RC\omega)^2}$$

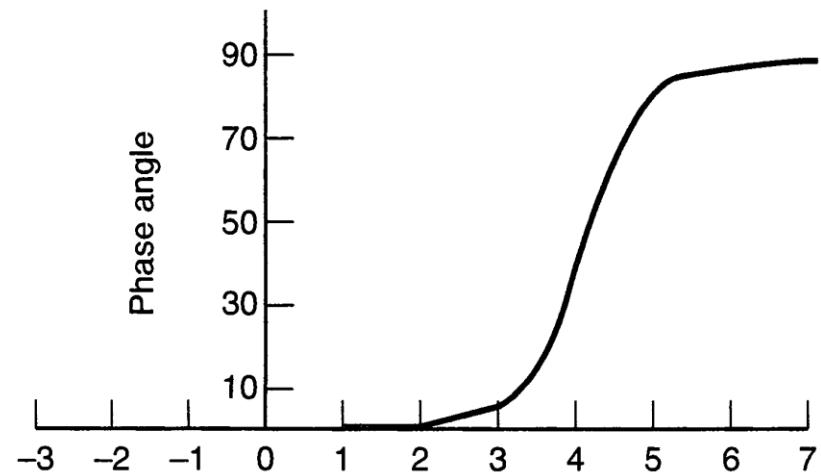
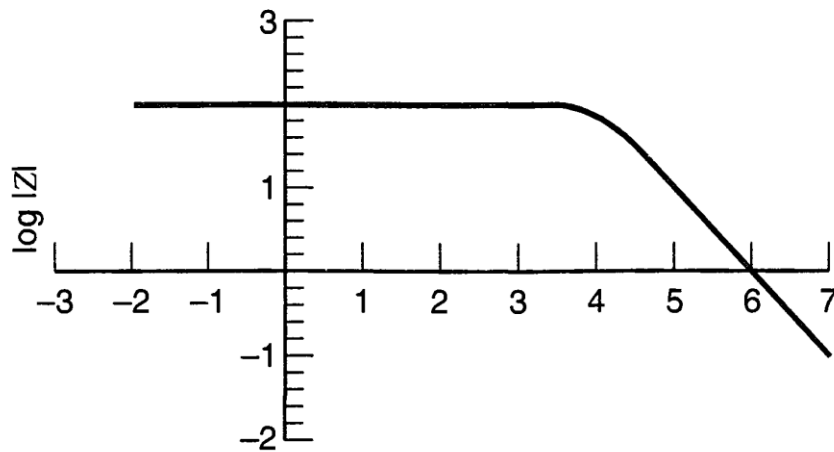
I) Fundamentals of EIS

5) Representing EIS data

$$Z = \frac{R}{1 + (RC\omega)^2} - j \frac{R^2 C \omega}{1 + (RC\omega)^2}$$



Bode Plots



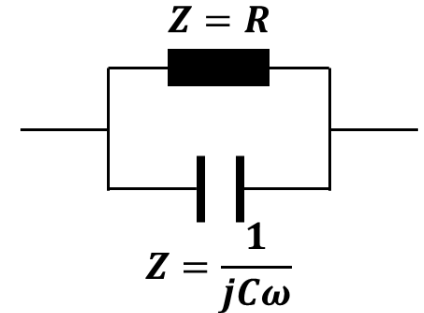
$\log \text{frequency}$

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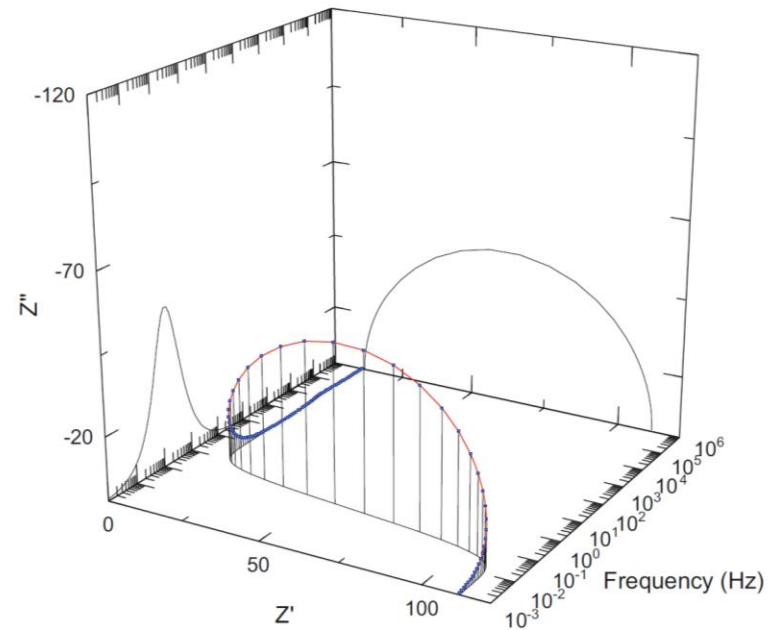
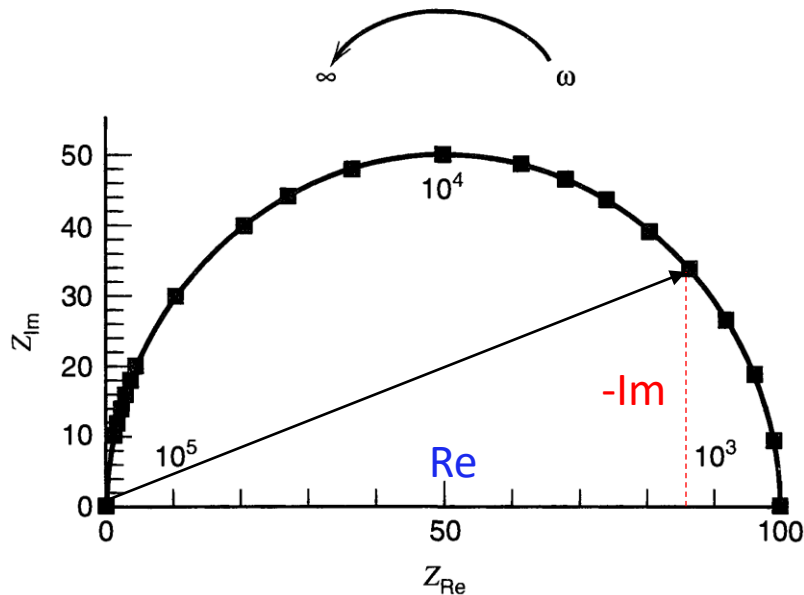
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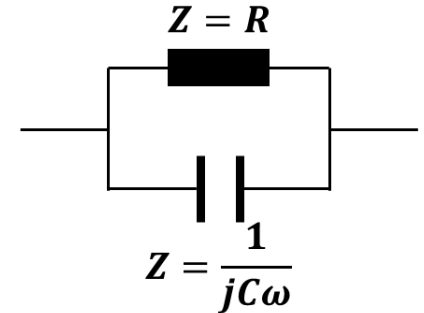
Nyquist Plots



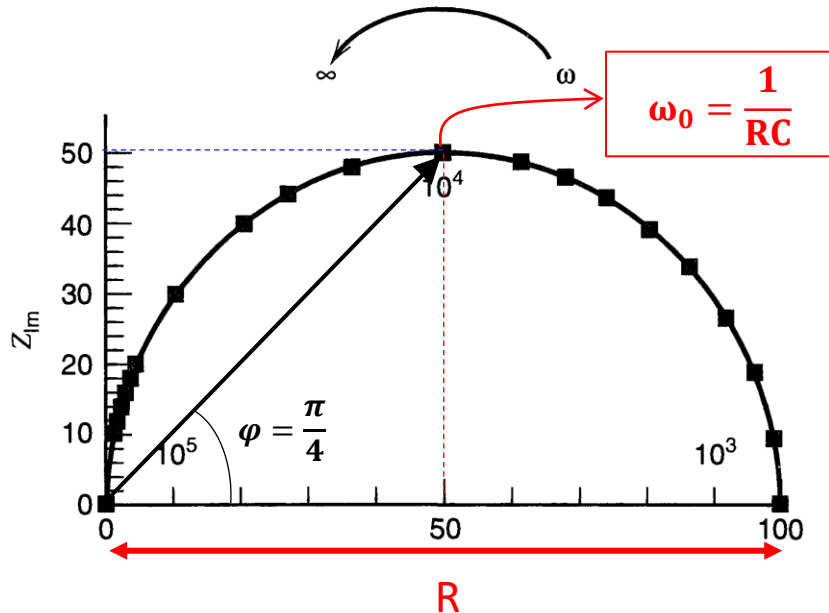
I) Fundamentals of EIS

6) Treatment of a ZARC element

$$Z = \frac{R}{1 + (RC\omega)^2} - j \frac{R^2 C \omega}{1 + (RC\omega)^2}$$



Nyquist Plots



Distinctive values:

$$Z(\omega \mapsto \infty) = 0$$

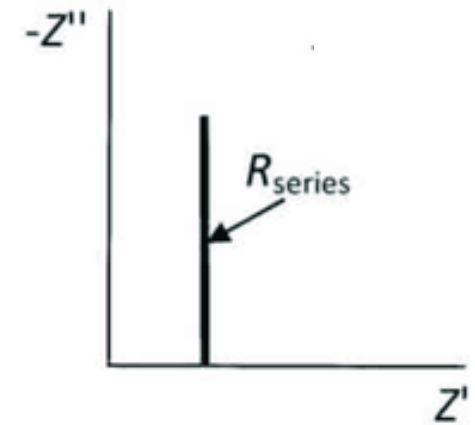
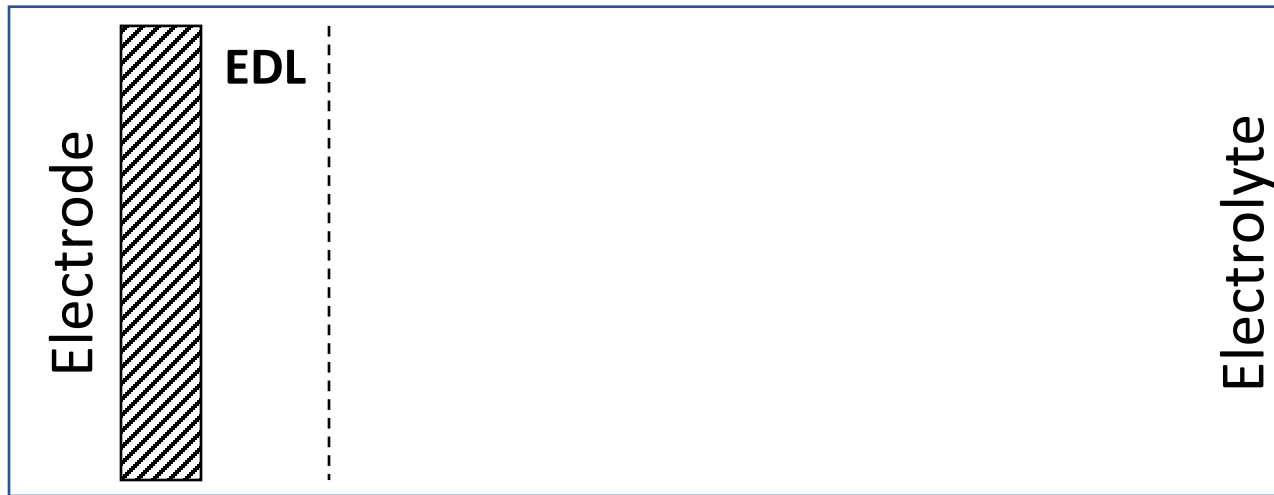
$$Z(\omega \mapsto 0) = Z_{Re} = R$$

$$Z_{Re}(\omega_0) = -Z_{Im}(\omega_0) \Rightarrow \omega_0 = \frac{1}{RC}$$

I) Fundamentals of EIS

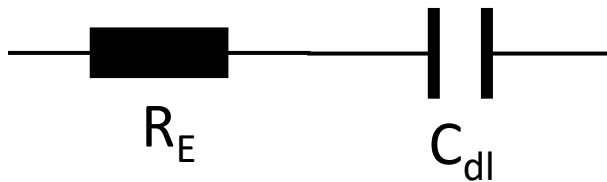
7) Faradaic reaction: the Randles circuit

Ideally polarisable electrode



Electrolyte resistance R_E

Electrical double layer capacitance: C_{dl}

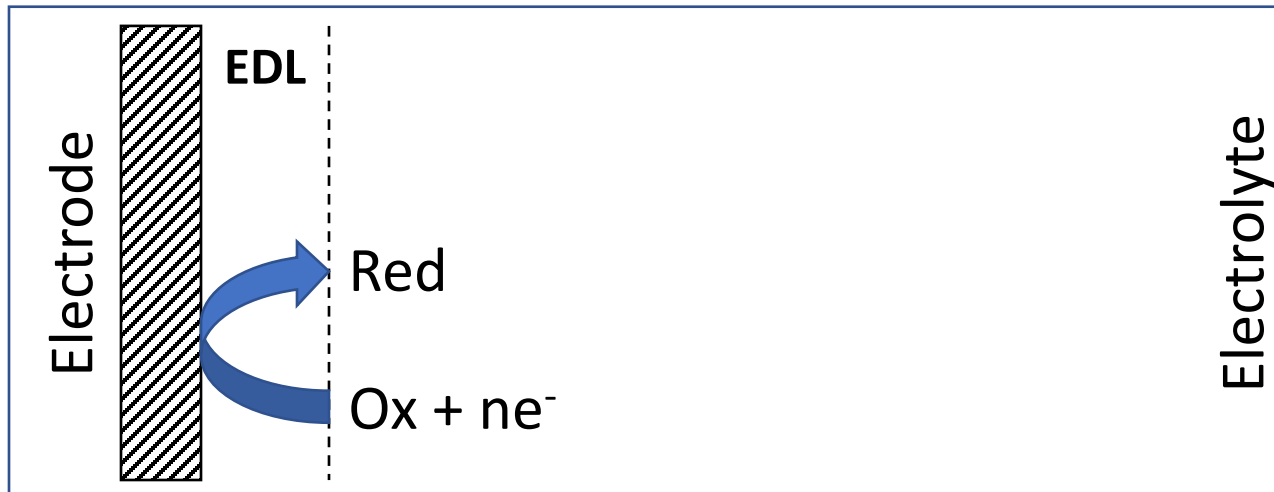


$$Z = R_E - j \frac{1}{C_{dl}\omega}$$

I) Fundamentals of EIS

7) Faradaic reaction: the Randles circuit

Charge transfer reaction

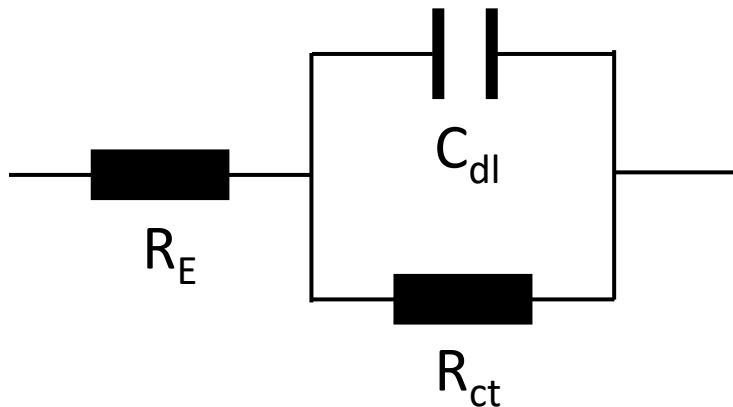


Prerequisites:

The system is linear

Ox and Red do not modify the electrode

High frequency \leftrightarrow no mass transfer limitation



Electrolyte resistance R_E

Electrical double layer capacitance: C_{dl}

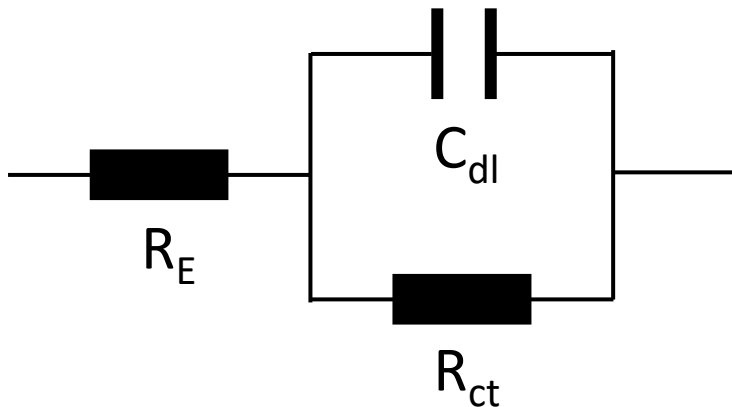
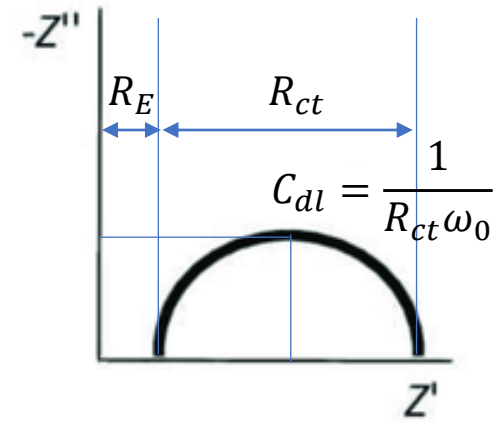
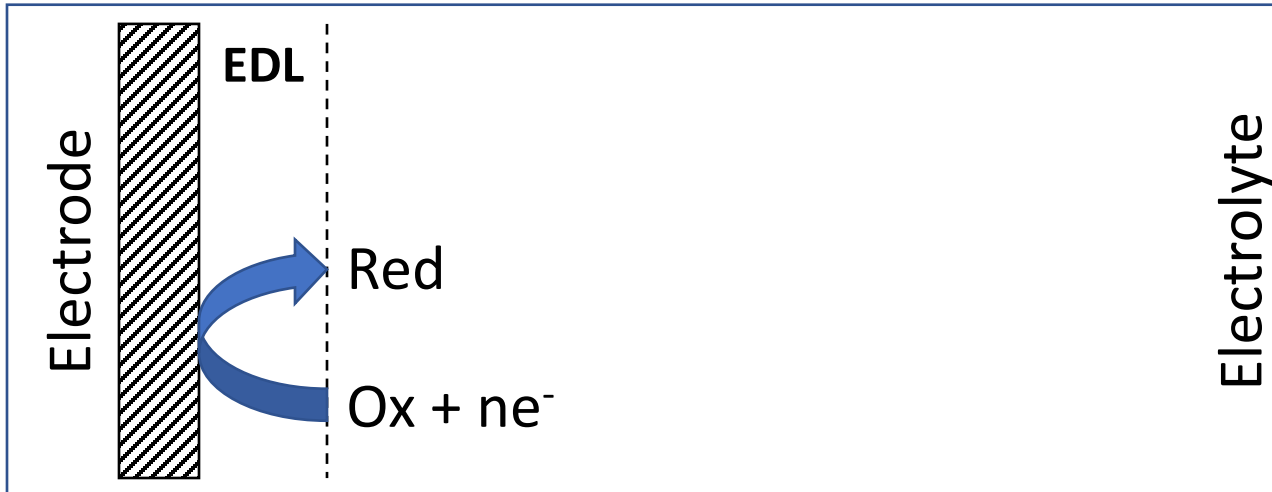
Charge transfer resistance R_{ct}

$$Z = R_E + \frac{R_{ct}}{1 + (R_{ct}C_{dl}\omega)^2} - j \frac{R_{ct}^2 C_{dl}\omega}{1 + (R_{ct}C_{dl}\omega)^2}$$

I) Fundamentals of EIS

7) Faradaic reaction: the Randles circuit

Charge transfer reaction



Electrolyte resistance R_E

Electrical double layer capacitance: C_{dl}

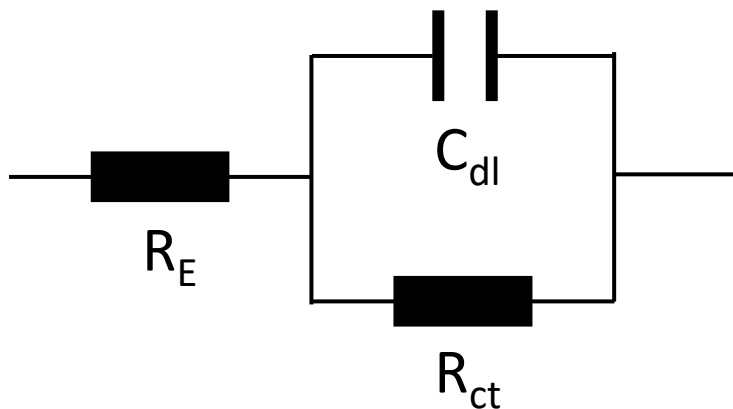
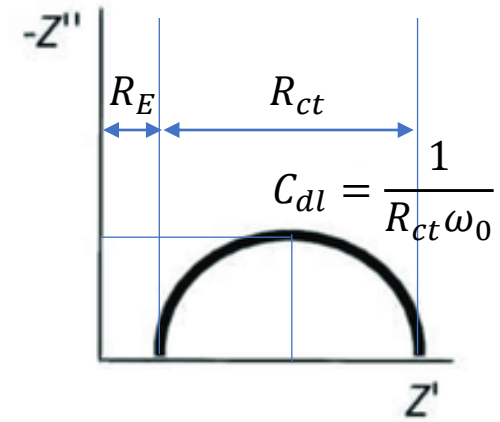
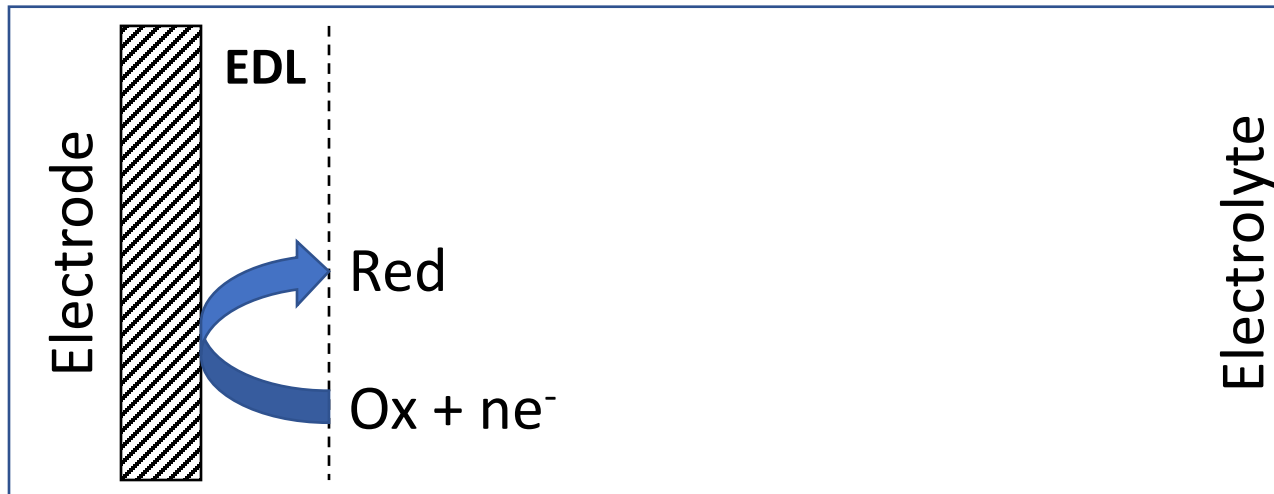
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I) Fundamentals of EIS

7) Faradaic reaction: the Randles circuit

Charge transfer reaction



Low η : linearized Butler-Volmer $i = i_0 \cdot \frac{zF\eta}{RT}$

hence $R_{ct} = \frac{RT}{zFi_0} \Leftrightarrow i_0 = \frac{RT}{zFR_{ct}}$

$$i_0 = zF[Ox]\beta_{Red} \cdot \exp\left(\frac{-\Delta G_{\chi, Red}^*}{RT}\right) \cdot \exp\left(\frac{-\alpha zF\Delta\phi_{eq}}{RT}\right)$$

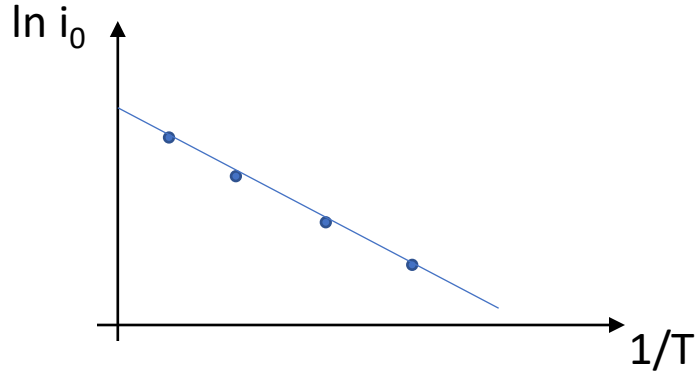
$$= zF[Red]\beta_{Ox} \cdot \exp\left(\frac{-\Delta G_{\chi, Ox}^*}{RT}\right) \cdot \exp\left(\frac{(1-\alpha)zF\Delta\phi_{eq}}{RT}\right)$$

I) Fundamentals of EIS

7) Faradaic reaction: the Randles circuit

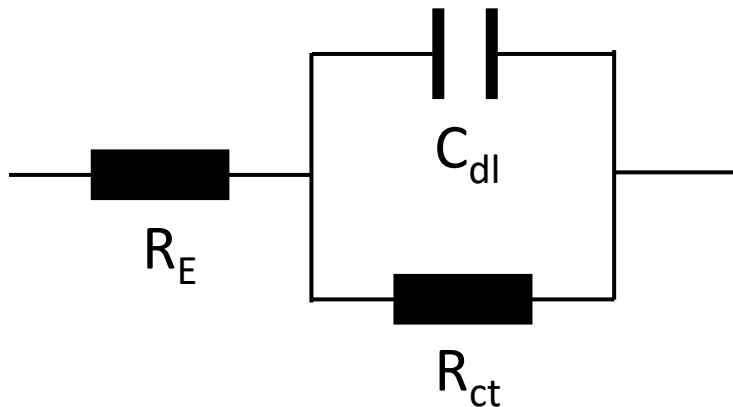
Charge transfer reaction

Measure R_{ct} at different temperatures: Arrhenius plot



$$\ln i_0 = -\frac{\Delta G_{\chi, Red}^* + \alpha zF \Delta \phi_{eq}}{RT} + \ln(zF[Ox]\beta_{Red})$$

$$\ln i_0 = -\frac{\Delta G_{\chi, Ox}^* + (1 - \alpha)zF \Delta \phi_{eq}}{RT} + \ln(zF[Red]\beta_{Ox})$$



Low η : linearized Butler-Volmer $i = i_0 \cdot \frac{zF\eta}{RT}$

$$\text{hence } R_{ct} = \frac{RT}{zFi_0} \Leftrightarrow i_0 = \frac{RT}{zFR_{ct}}$$

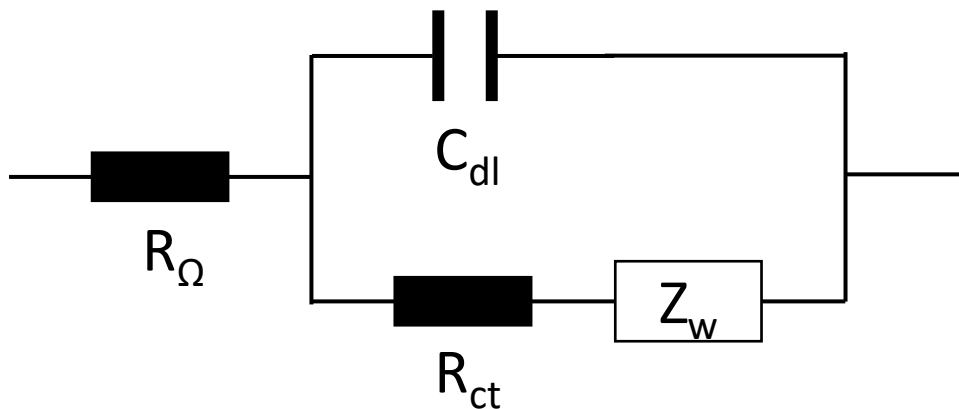
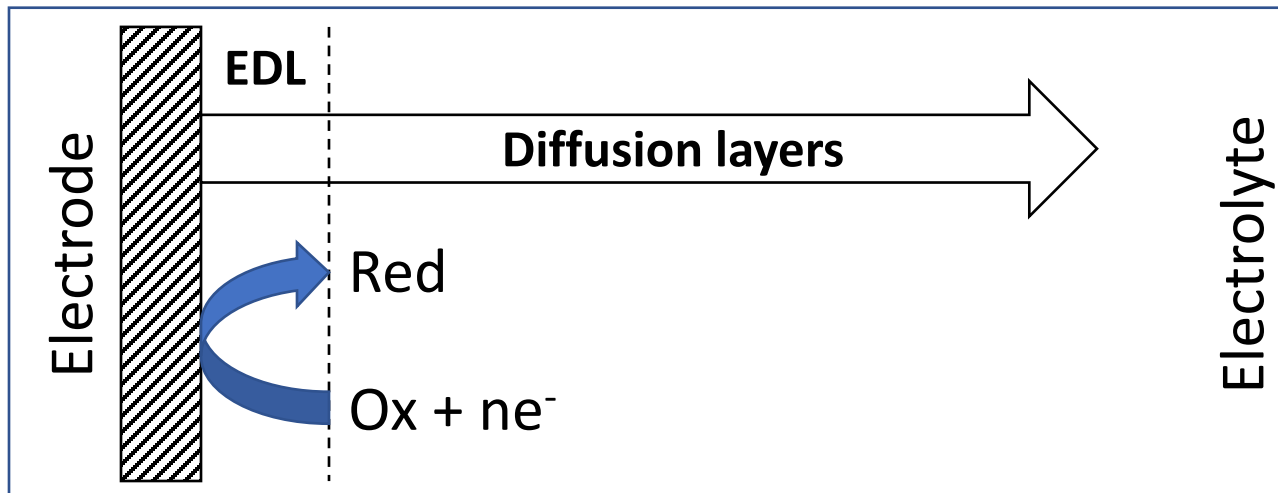
$$i_0 = zF[Ox]\beta_{Red} \cdot \exp\left(\frac{-\Delta G_{\chi, Red}^*}{RT}\right) \cdot \exp\left(\frac{-\alpha zF \Delta \phi_{eq}}{RT}\right)$$

$$= zF[Red]\beta_{Ox} \cdot \exp\left(\frac{-\Delta G_{\chi, Ox}^*}{RT}\right) \cdot \exp\left(\frac{(1-\alpha)zF \Delta \phi_{eq}}{RT}\right)$$

I) Fundamentals of EIS

7) Faradaic reaction: the Randles circuit

$[Ox]^{el}$ and $[Red]^{el}$ are time dependent



Ohmic resistance of the circuit R_{Ω}

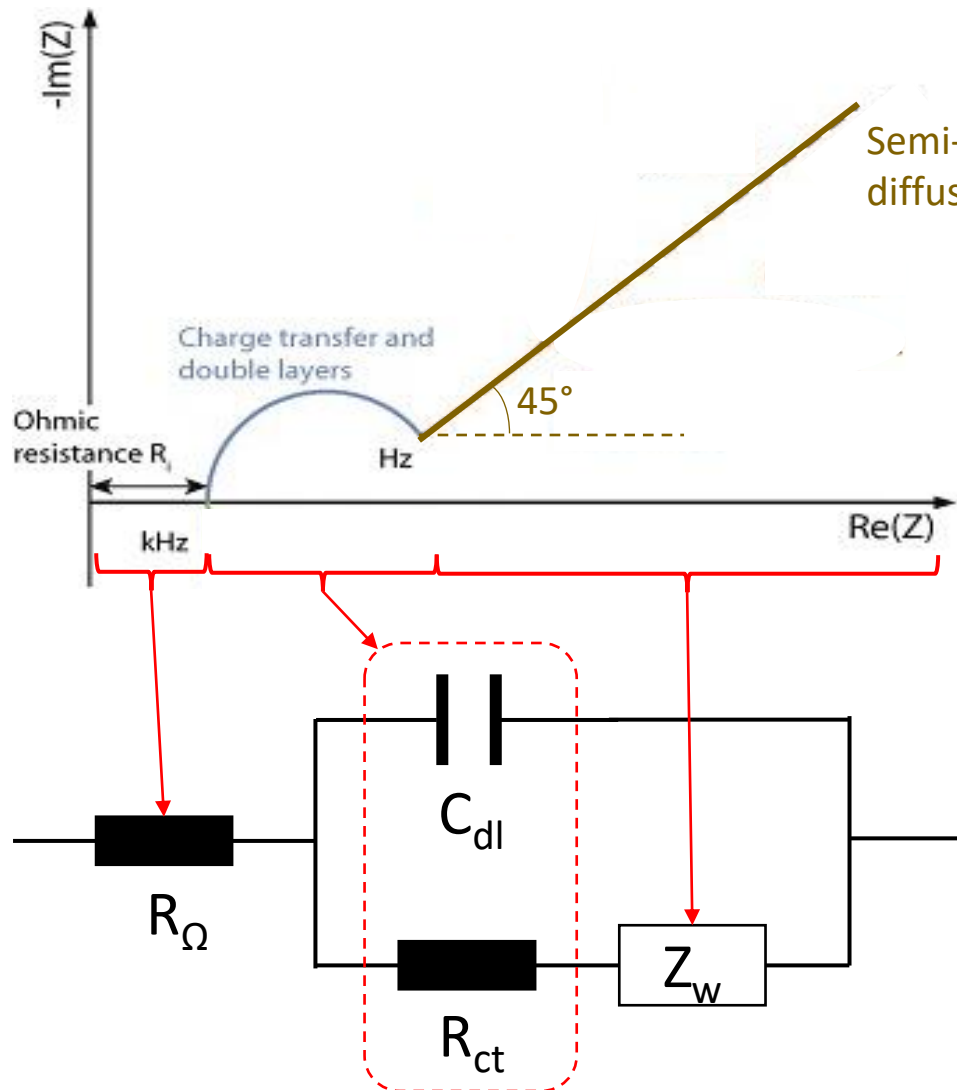
Electrical double layer capacitance: C_{dl}

Charge transfer resistance R_{ct}

Warburg Impedance Z_w for diffusion

1) Fundamentals of EIS

7) Faradaic reaction: the Randles circuit



Ohmic resistance of the circuit R_Ω

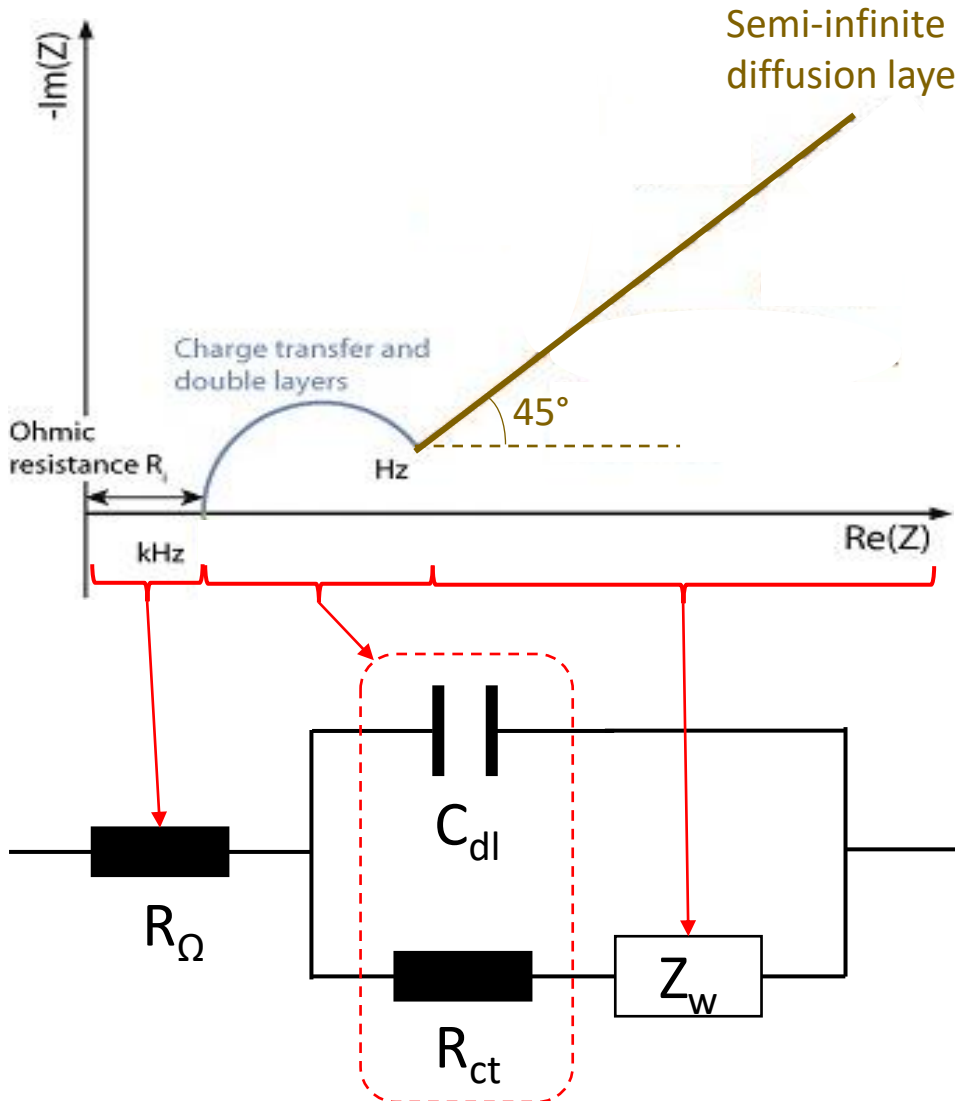
Electrical double layer capacitance: C_{dl}

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Warburg Impedance Z_w for diffusion

I) Fundamentals of EIS

7) Faradaic reaction: the Randles circuit



Butler-Volmer with mass limitation:

$$\bar{i} = i_0 \left(\frac{\bar{C}_R}{C_{R^0}} \exp\left(\frac{(1-\alpha)zF\bar{\eta}}{RT}\right) - \frac{\bar{C}_O}{C_{O^0}} \exp\left(\frac{-\alpha zF\bar{\eta}}{RT}\right) \right)$$

Fick's 2nd law of diffusion:

$$\frac{\partial C_i}{\partial t} = D_i \frac{\partial^2 C_i}{\partial x^2}$$

with $C_i(t) = C_i^{DC} + C_i^{AC} \exp^{j(\omega t + \phi_i)}$

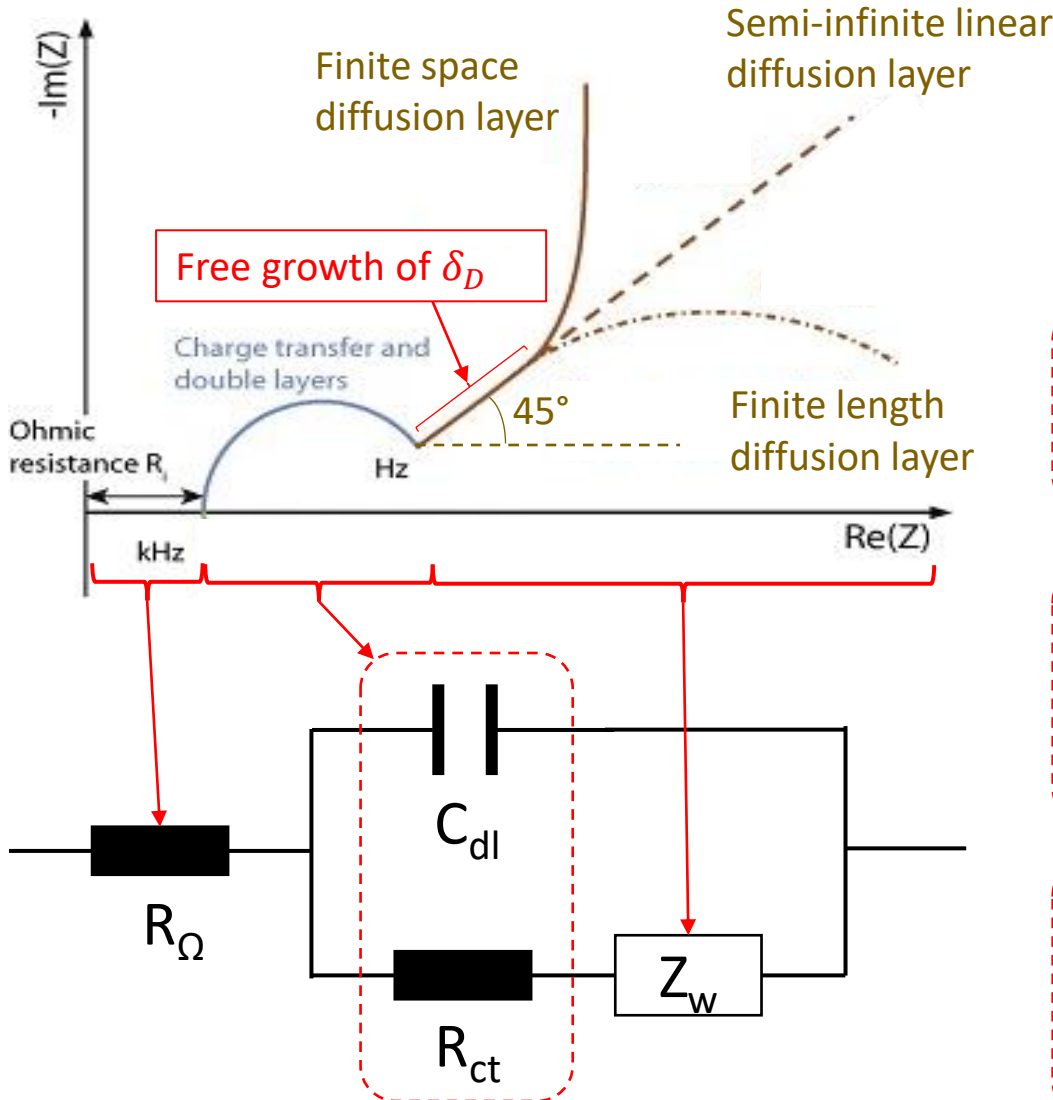
Boundary conditions:

$$\left[\begin{array}{l} x \mapsto 0 \Rightarrow \bar{i} = zFD_i \frac{d\bar{C}_i}{dx} \\ x \mapsto \infty \Rightarrow C_i = C_i^0 \\ D_R \frac{dC_R}{dx} = -D_O \frac{dC_O}{dx} \end{array} \right] \quad \bar{C}_i = \frac{\bar{i}}{zF\sqrt{j\omega D_i}} \exp^{-\sqrt{\frac{j\omega}{D_i}}x}$$

$$Z_w^{\infty} = \frac{\bar{E}}{\bar{i}} = \frac{1-j}{\sqrt{\omega}} \cdot \frac{RT}{\sqrt{2}(zF)^2} \left\{ \frac{1}{C_{O^0}\sqrt{D_O}} + \frac{1}{C_{R^0}\sqrt{D_R}} \right\}$$

I) Fundamentals of EIS

7) Faradaic reaction: the Randles circuit



Initial boundary conditions:

$$\left\{ \begin{array}{l} x \mapsto 0 \Rightarrow \tilde{i} = zFD_i \frac{d\tilde{C}_i}{dx} \\ D_R \frac{dC_R}{dx} = -D_O \frac{dC_O}{dx} \end{array} \right.$$

Semi-infinite δ_D : $x \mapsto \infty \Rightarrow C_i = C_i^0$

$$Z_W^\infty = \frac{1-j}{\sqrt{\omega}} \cdot \frac{RT}{\sqrt{2}(zF)^2} \left\{ \frac{1}{C_O^0 \sqrt{D_O}} + \frac{1}{C_R^0 \sqrt{D_R}} \right\}$$

Finite space δ_D : $x = l \Rightarrow \frac{dC_i}{dx} = 0$

$$Z_W^{FS} = \frac{1-j}{\sqrt{\omega}} \cdot \frac{RT}{\sqrt{2}(zF)^2} \left\{ \frac{\coth \sqrt{\frac{j\omega l}{D_O}}}{C_O^0 \sqrt{D_O}} + \frac{\coth \sqrt{\frac{j\omega l}{D_R}}}{C_R^0 \sqrt{D_R}} \right\}$$

Finite length δ_D : $x = l \Rightarrow \delta_D = \delta_D^{max}$

$$Z_W^{FS} = \frac{1-j}{\sqrt{\omega}} \cdot \frac{RT}{\sqrt{2}(zF)^2} \left\{ \frac{\tanh \sqrt{\frac{j\omega l}{D_O}}}{C_O^0 \sqrt{D_O}} + \frac{\tanh \sqrt{\frac{j\omega l}{D_R}}}{C_R^0 \sqrt{D_R}} \right\}$$

I) Fundamentals of EIS

8) Constant phase element (CPE)

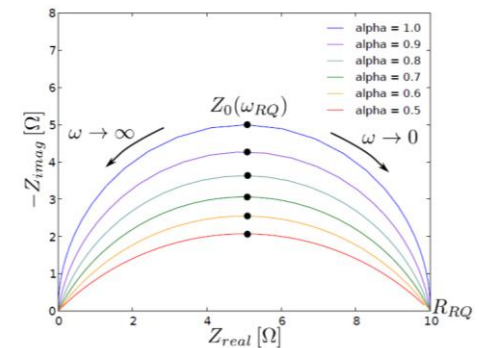
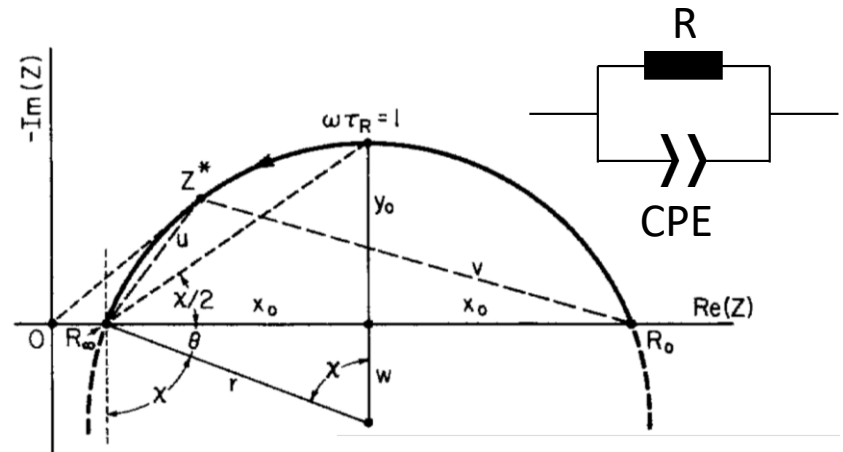
The Electrical Double Layer is not an ideal capacitor

There are distributed elements at the electrode/electrolyte interphase

$$Z_Q = \frac{1}{Q(j\omega)^\alpha} \quad \text{with } 0 \leq \alpha \leq 1$$

$$Z_{ZARC} = \frac{R}{(1 + RQ(j\omega)^\alpha)}$$

$$\alpha = \frac{2\chi}{\Pi}$$



Dispersion of the relaxation frequency \nearrow as $\alpha \searrow$

I) Fundamentals of EIS

8) Constant phase element (CPE)

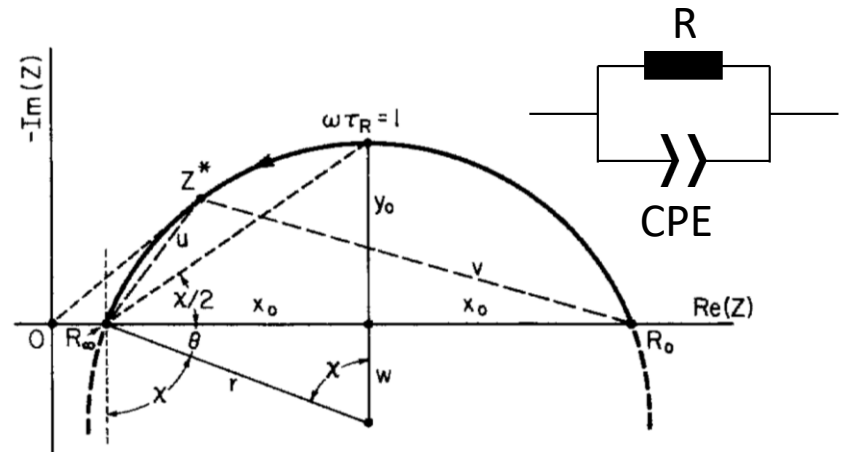
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$$\alpha = \frac{2\chi}{\Pi}$$



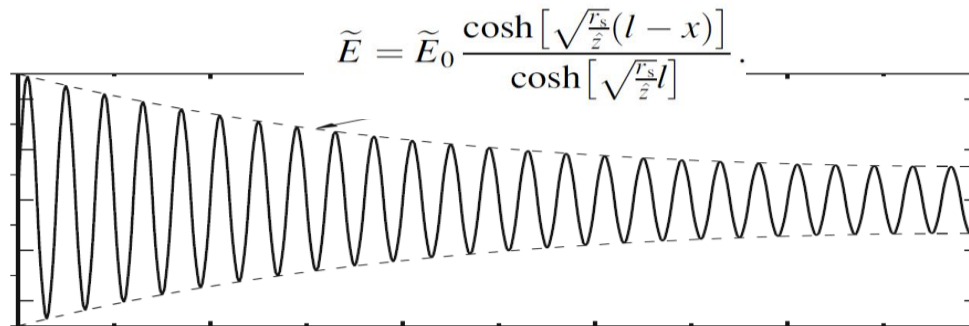
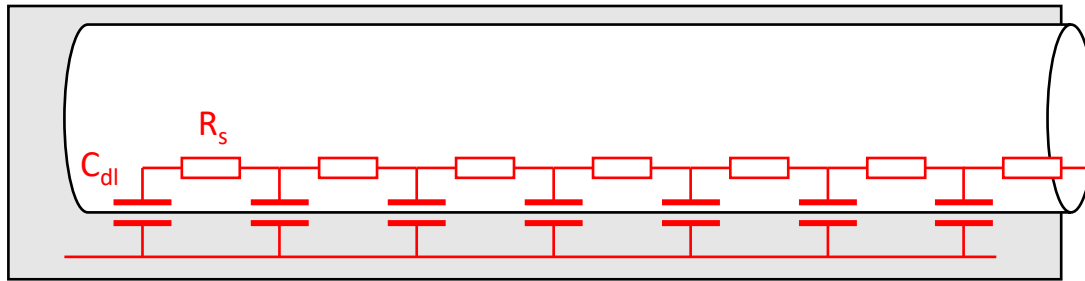
$$Re = \frac{R + \omega^\alpha R^2 Q \cos\left(\frac{\alpha}{2}\pi\right)}{1 + 2\omega^\alpha RQ \cos\left(\frac{\alpha}{2}\pi\right) + (\omega^\alpha RQ)^2}$$

$$Im = \frac{-\omega^\alpha R^2 Q \sin\left(\frac{\alpha}{2}\pi\right)}{1 + 2\omega^\alpha RQ \cos\left(\frac{\alpha}{2}\pi\right) + (\omega^\alpha RQ)^2}$$

I) Fundamentals of EIS

8) Transmission lines: De Levie model for porous media

A conductive pore in an ionic conductor



$$\hat{Z}_{\text{pore}} = \frac{R_{\Omega,p}}{\Lambda^{1/2}} \coth(\Lambda^{1/2})$$

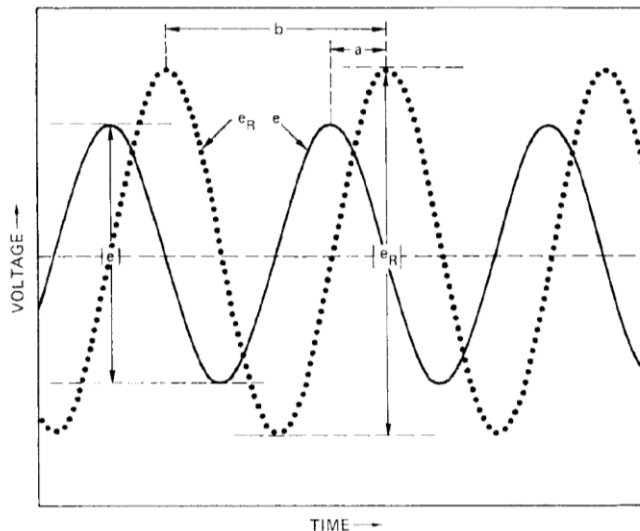
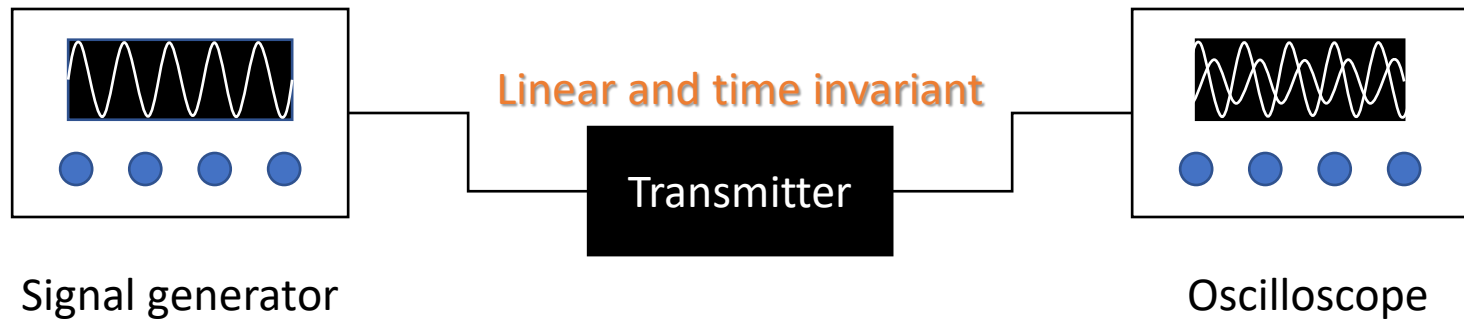
$$\Lambda = \frac{r_s l^2}{z} = \frac{2\rho_s l^2}{r \hat{Z}_{\text{el}}} = \frac{2\rho_s l^2 j\omega C_{\text{dl}}}{r}$$

I) Fundamentals of EIS

element	impedance	process
(R) serial resistor	R	contact and electrolyte resistance
(RQ) resistor and a constant phase element in parallel	$\frac{R}{1+RQ_0(j\omega)^n}$	double layer capacitance and charge-transfer reactions in parallel
(G) Gerischer element	$\frac{R_{\text{chem}}}{\sqrt{1+j\omega t_{\text{chem}}}}$	combination of electrochemical charge-transfer and diffusion
(W) Warburg element	$R_w \frac{\tanh[(j\omega T)^\alpha]}{(j\omega T)^\alpha}$	diffusion processes
(TLM) transition line model	$\frac{R_{\text{el}}R_{\text{ion,L}}}{R_{\text{el}}+R_{\text{ion,L}}} \left(L + \frac{2\lambda}{\sinh(L/\lambda)} \right) + \lambda_{TLM} \frac{R_{\text{el}}^2+R_{\text{ion,L}}^2}{R_{\text{el}}+R_{\text{ion,L}}} \coth \left(\frac{L}{\lambda_{TLM}} \right)$	porous electrode electrochemistry

II) Practical aspects, analyses, and applications

1) Impedance measurements



Measure f , i_{DC} , E_{DC} , Δi , and ΔE

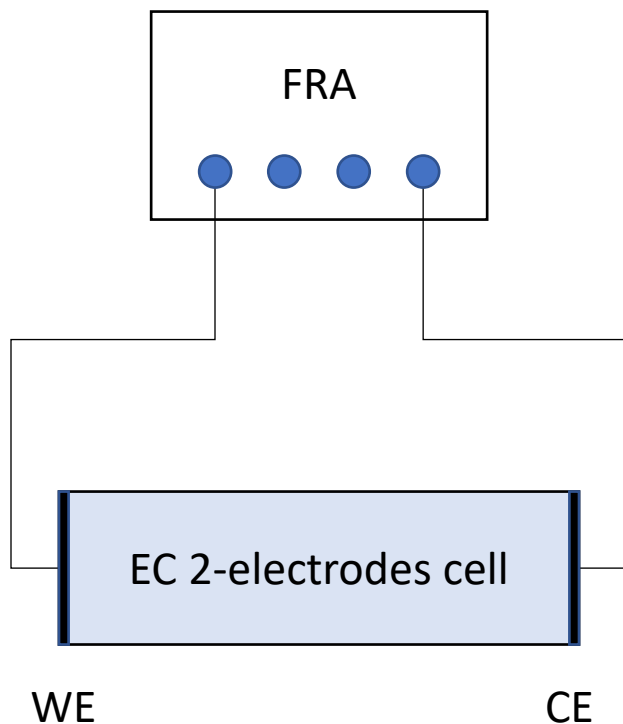
Visual control of harmonics

To ensure the system linearity, we typically target $\Delta E \approx 10 \text{ mV}$

II) Practical aspects, analyses, and applications

1) Impedance measurements

Potentiostat/galvanostat
+ Frequency Response Analyzer



Possible to measure a 2-electrodes EC-cell

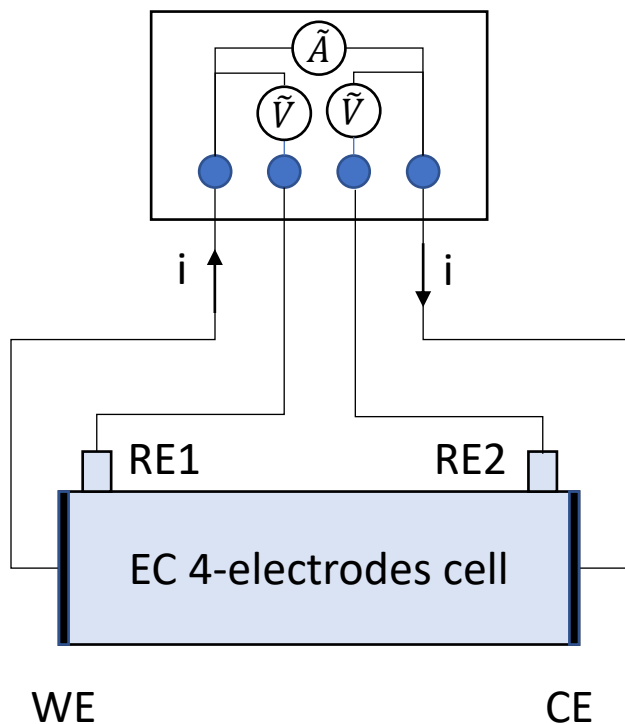
Complex equivalent circuit

- difficult to interpret
- deconvolute processes
- parametric study necessary

II) Practical aspects, analyses, and applications

1) Impedance measurements

Potentiostat/galvanostat
+ Frequency Response Analyzer



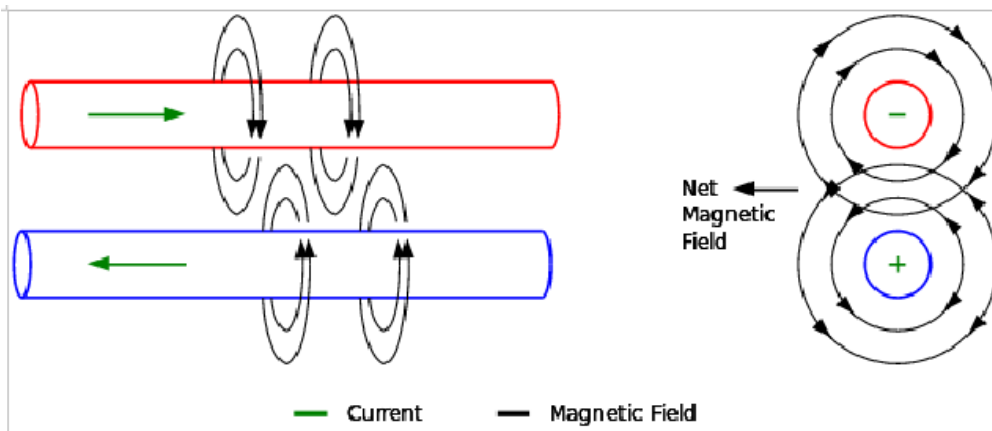
4-electrodes EC-cell easier to interpret

Separation of cathodic and anodic elements

II) Practical aspects, analyses, and applications

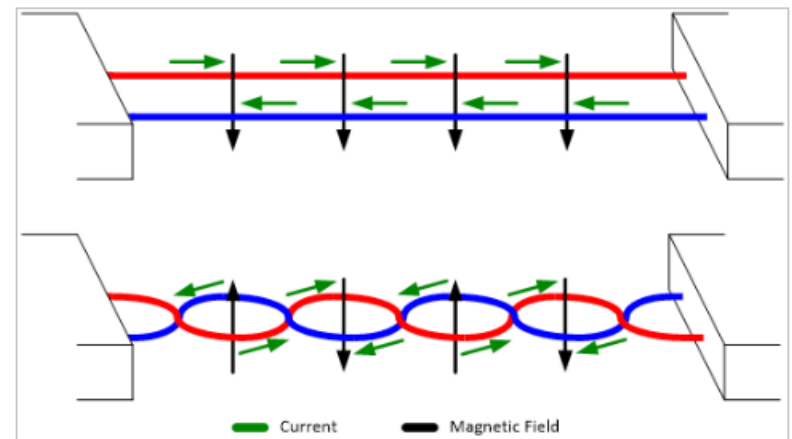
1) Impedance measurements

Suppressing the wire inductance



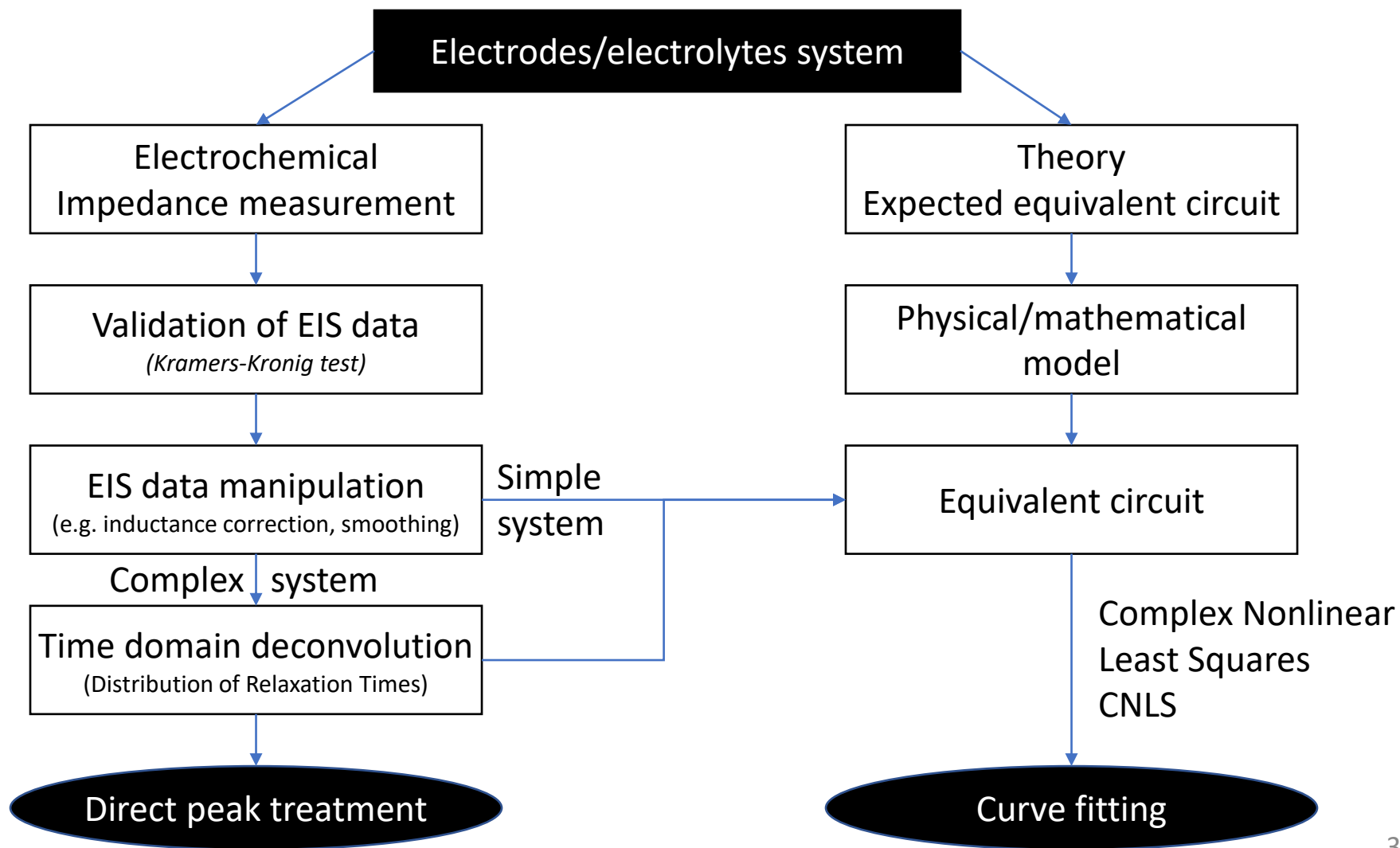
Wire inductance results from opposite currents flowing through parallel wires

Compensation of inductive currents by uniform twisting of the wires



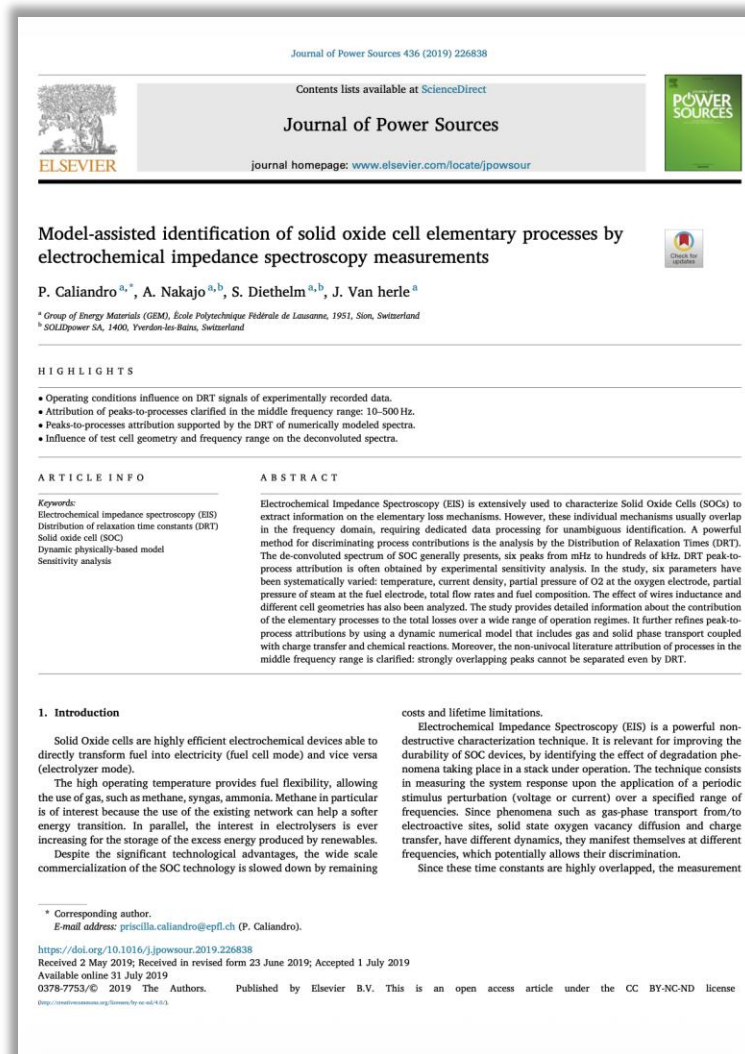
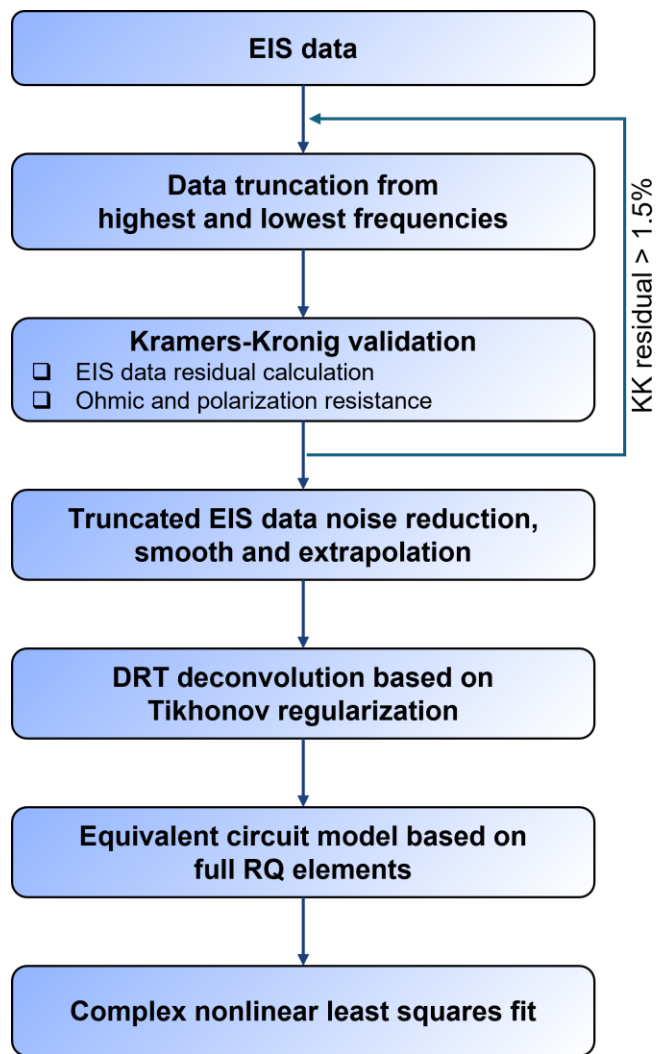
II) Practical aspects, analyses, and applications

2) Data treatment



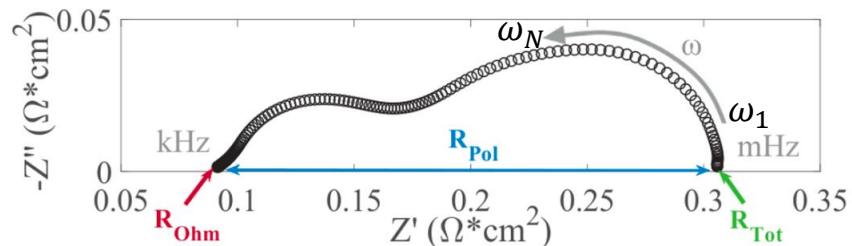
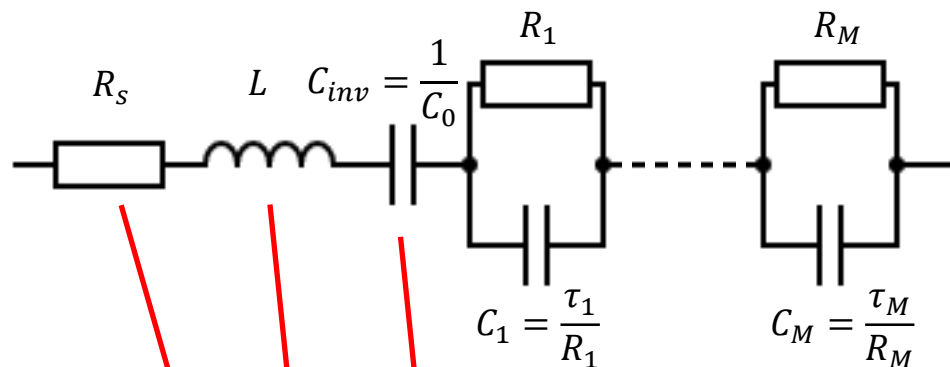
II) Practical aspects, analyses, and applications

2) Data treatment



II) Practical aspects, analyses, and applications

Kramers-Kronig Transform (Linearity test)



$$Z_{fit,i} = R_s + j\omega_i L + \frac{C_{inv}}{j\omega_i} + \sum_{k=1}^M \frac{R_k}{1 + (\omega_i \tau_k)^2} = \underbrace{\left(R_s + \sum_{k=1}^M \frac{R_k}{1 + (\omega_i \tau_k)^2} \right)}_{Re_{fit,i}} + j \cdot \underbrace{\left(\omega_i L - \frac{C_{inv}}{\omega_i} - \sum_{k=1}^M \frac{R_k \omega_i \tau_k}{1 + (\omega_i \tau_k)^2} \right)}_{Im_{fit,i}}$$

$$Z_{fit} = \sum_{i=1}^N Z_{fit,i} = \sum_{i=1}^N (Re_{fit,i} + j \cdot Im_{fit,i})$$

$$Re_{fit} = \sum_{i=1}^N \left(R_s + \sum_{k=1}^M \frac{R_k}{1 + (\omega_i \tau_k)^2} \right)$$

$$Im_{fit} = \sum_{i=1}^N \left(\omega_i L - \frac{C_{inv}}{\omega_i} - \sum_{k=1}^M \frac{R_k \omega_i \tau_k}{1 + (\omega_i \tau_k)^2} \right)$$

$$\tau_k = R_k C_k = 10^{\log_{10}(\tau_{min}) + \frac{k-1}{M-1} \cdot [\log_{10}(\tau_{max}) - \log_{10}(\tau_{min})]}$$

$$S = \sum_{i=1}^N \lambda_i [(Re_i - Re_{fit,i})^2 + (Im_i - Im_{fit,i})^2]$$

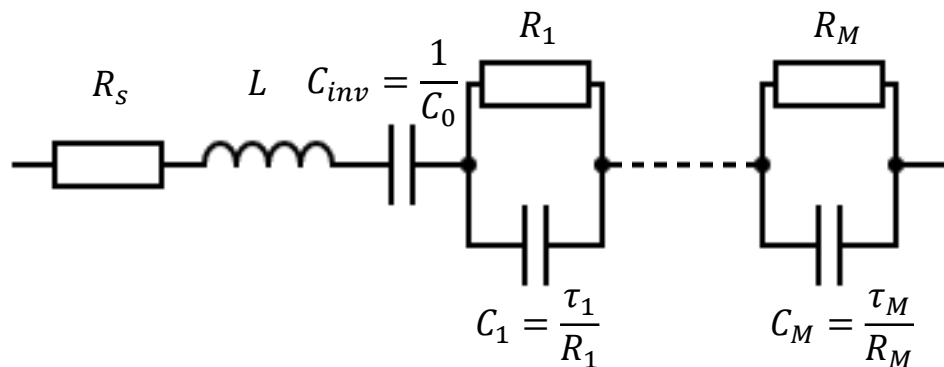
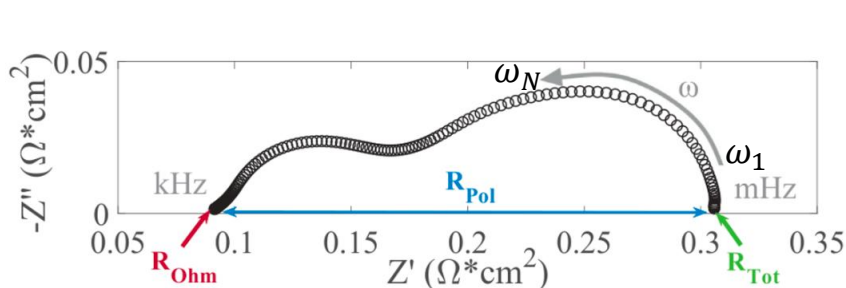
$$\text{Weight: } \lambda_i = \frac{1}{|Z_i|^2} = \frac{1}{Re_i^2 + Im_i^2}$$

$$\text{Objective: } \text{argmin} S \rightarrow \nabla S = 0 \rightarrow \frac{\delta S}{\delta R_j} = 0$$

$$\text{Unknown parameters: } X = [R_s, L, C_{inv}, R_1, R_2, \dots, R_M]^T$$

II) Practical aspects, analyses, and applications

Kramers-Kronig Transform (Linearity test)



$$R_s \text{ as an example with } S = \sum_{i=1}^N \lambda_i [(Re_i - Re_{fit,i})^2 + (Im_i - Im_{fit,i})^2]:$$

$$\frac{\delta S}{\delta R_s} = \sum_{i=1}^N \left[\lambda_i \cdot 2(Re_i - Re_{fit,i}) \cdot \frac{\delta Re_{fit,i}}{\delta R_s} + \lambda_i \cdot 2(Im_i - Im_{fit,i}) \cdot \frac{\delta Im_{fit,i}}{\delta R_s} \right]$$

$$\because Re_{fit,i} = R_s + \sum_{k=1}^M \frac{R_k}{1 + (\omega_i \tau_k)^2}, \quad Im_{fit,i} = \omega_i L - \frac{C_{inv}}{\omega_i} - \sum_{k=1}^M \frac{R_k \omega_i \tau_k}{1 + (\omega_i \tau_k)^2}$$

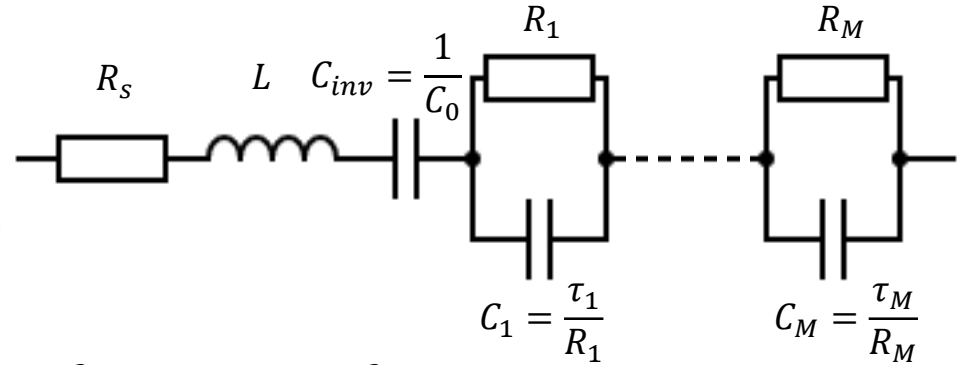
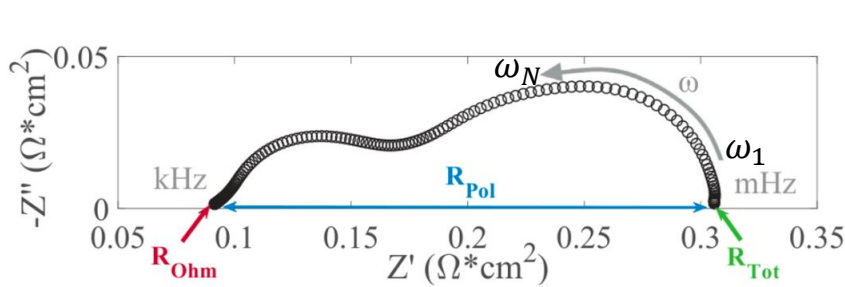
$$\therefore \frac{\delta Re_{fit,i}}{\delta R_s} = 1, \quad \frac{\delta Im_{fit,i}}{\delta R_s} = 0$$

$$\therefore \frac{\delta S}{\delta R_s} = \sum_{i=1}^N \lambda_i \cdot 2(Re_i - Re_{fit,i}) = 0$$

$$\therefore \sum_{i=1}^N \lambda_i Re_{fit,i} = \sum_{i=1}^N \lambda_i Re_i \Leftrightarrow R_s \sum_{i=1}^N \lambda_i + \sum_{k=1}^M R_k \sum_{i=1}^N \frac{\lambda_i}{1 + (\omega_i \tau_k)^2} = \sum_{i=1}^N \lambda_i Re_i \Leftrightarrow AX = b$$

II) Practical aspects, analyses, and applications

Kramers-Kronig Transform (Linearity test)



L as an example with $S = \sum_{i=1}^N \lambda_i [(Re_i - Re_{fit,i})^2 + (Im_i - Im_{fit,i})^2]$:

$$\frac{\delta S}{\delta L} = \sum_{i=1}^N \left[\lambda_i \cdot 2(Re_i - Re_{fit,i}) \cdot \frac{\delta Re_{fit,i}}{\delta L} + \lambda_i \cdot 2(Im_i - Im_{fit,i}) \cdot \frac{\delta Im_{fit,i}}{\delta L} \right]$$

$$\because Re_{fit,i} = R_s + \sum_{k=1}^M \frac{R_k}{1 + (\omega_i \tau_k)^2}, \quad Im_{fit,i} = \omega_i L - \frac{C_{inv}}{\omega_i} - \sum_{k=1}^M \frac{R_k \omega_i \tau_k}{1 + (\omega_i \tau_k)^2}$$

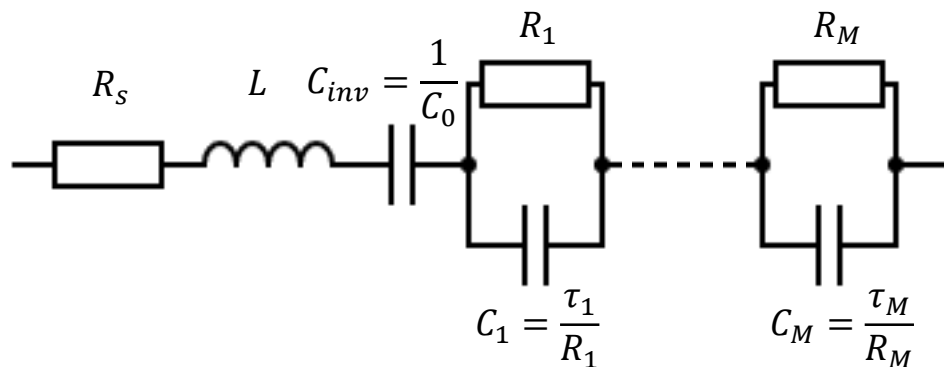
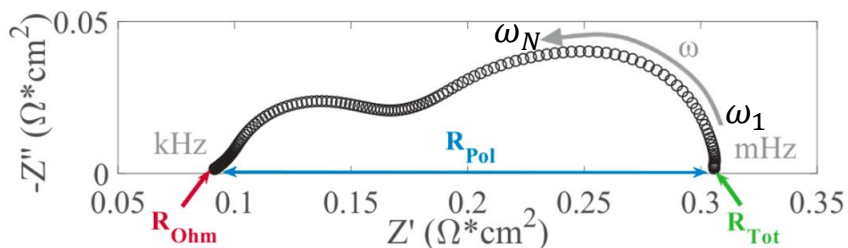
$$\because \frac{\delta Re_{fit,i}}{\delta L} = 0, \quad \frac{\delta Im_{fit,i}}{\delta L} = \omega_i$$

$$\because \frac{\delta S}{\delta L} = \sum_{i=1}^N \lambda_i \cdot 2\omega_i (Im_i - Im_{fit,i}) = 0$$

$$\because \sum_{i=1}^N \lambda_i \omega_i Im_{fit,i} = \sum_{i=1}^N \lambda_i \omega_i Im_i \Leftrightarrow \sum_{i=1}^N \lambda_i \omega_i L - \sum_{i=1}^N \frac{\lambda_i}{C_{inv}} - \sum_{k=1}^M R_k \sum_{i=1N} \frac{\omega_i^2 \tau_k}{1 + (\omega_i \tau_k)^2} = \sum_{i=1}^N \lambda_i \omega_i Im_i \Leftrightarrow AX = b$$

II) Practical aspects, analyses, and applications

Kramers-Kronig Transform (Linearity test)



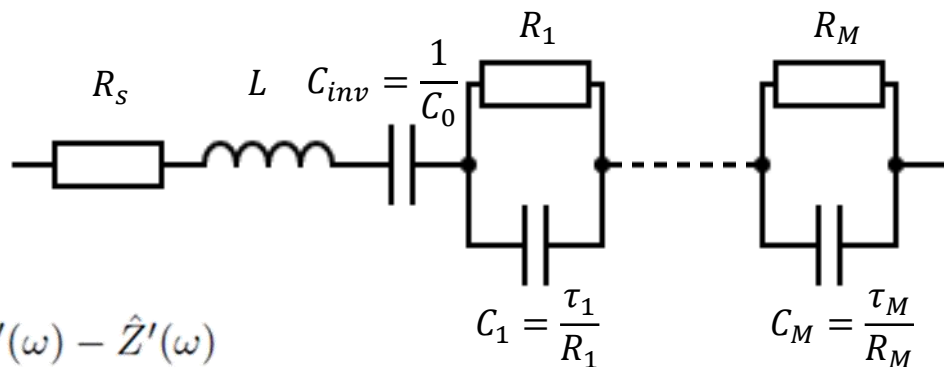
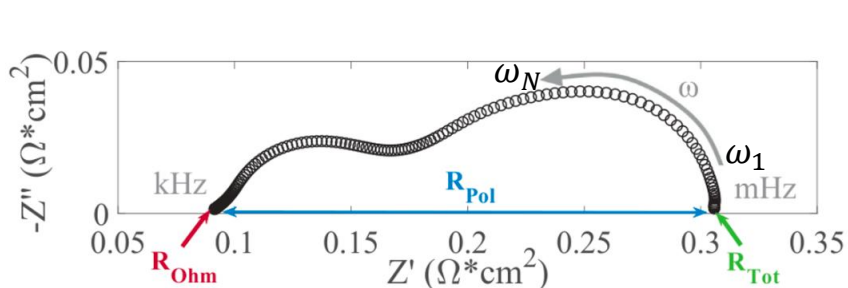
Linear algebra $\mathbf{AX} = \mathbf{b}$:

$$\begin{pmatrix}
 \sum_{i=1}^N \lambda_i & 0 & 0 & \dots & \sum_{i=1}^N \frac{\lambda_i}{1 + (\omega_i \tau_M)^2} \\
 0 & \sum_{i=1}^N \lambda_i \omega_i^2 & -\sum_{i=1}^N \lambda_i & \dots & -\sum_{i=1}^N \frac{\lambda_i \omega_i^2 \tau_M}{1 + (\omega_i \tau_M)^2} \\
 \sum_{i=1}^N \frac{\lambda_i}{1 + (\omega_i \tau_1)^2} & -\sum_{i=1}^N \lambda_i & \sum_{i=1}^N \frac{\lambda_i}{\omega_i^2} & \dots & -\sum_{i=1}^N \frac{\lambda_i \tau_M}{1 + (\omega_i \tau_M)^2} \\
 \vdots & \vdots & \vdots & \ddots & \vdots \\
 \sum_{i=1}^N \frac{\lambda_i}{1 + (\omega_i \tau_M)^2} & \sum_{i=1}^N \frac{\lambda_i \omega_i \tau_M}{1 + (\omega_i \tau_M)^2} & -\sum_{i=1}^N \frac{\lambda_i \tau_M}{\omega_i [1 + (\omega_i \tau_M)^2]} & \dots & \sum_{i=1}^N \frac{\lambda_i [1 + \omega_i \tau_M^2]}{\omega_i [1 + (\omega_i \tau_M)^2]^2}
 \end{pmatrix}
 \begin{pmatrix}
 R_s \\
 L \\
 C_{inv} \\
 \vdots \\
 R_M
 \end{pmatrix}
 =
 \begin{pmatrix}
 \sum_{i=1}^N \lambda_i Re_i \\
 \sum_{i=1}^N \lambda_i \omega_i Im_i \\
 \sum_{i=1}^N \frac{\lambda_i}{\omega_i} Im_i \\
 \vdots \\
 \sum_{i=1}^N \frac{\lambda_i (Re_i - \omega_i \tau_M Im_i)}{1 + (\omega_i \tau_M)^2}
 \end{pmatrix}$$

Singular value decomposition (SVD) to solve \mathbf{X} .

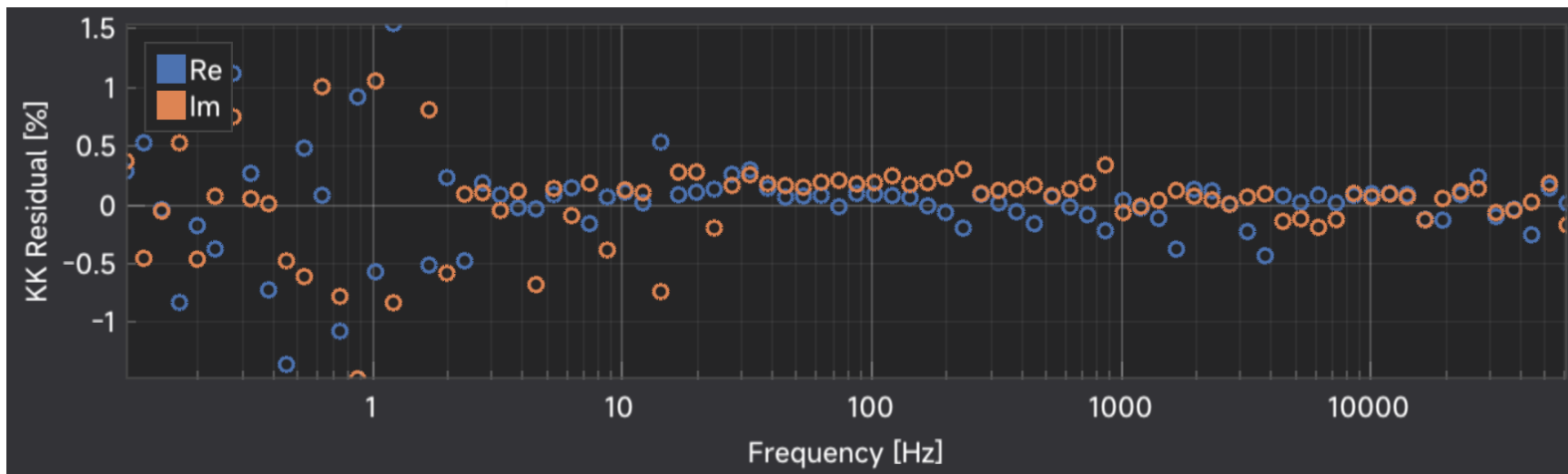
II) Practical aspects, analyses, and applications

Smooth, LC correction, extrapolation based on KK test



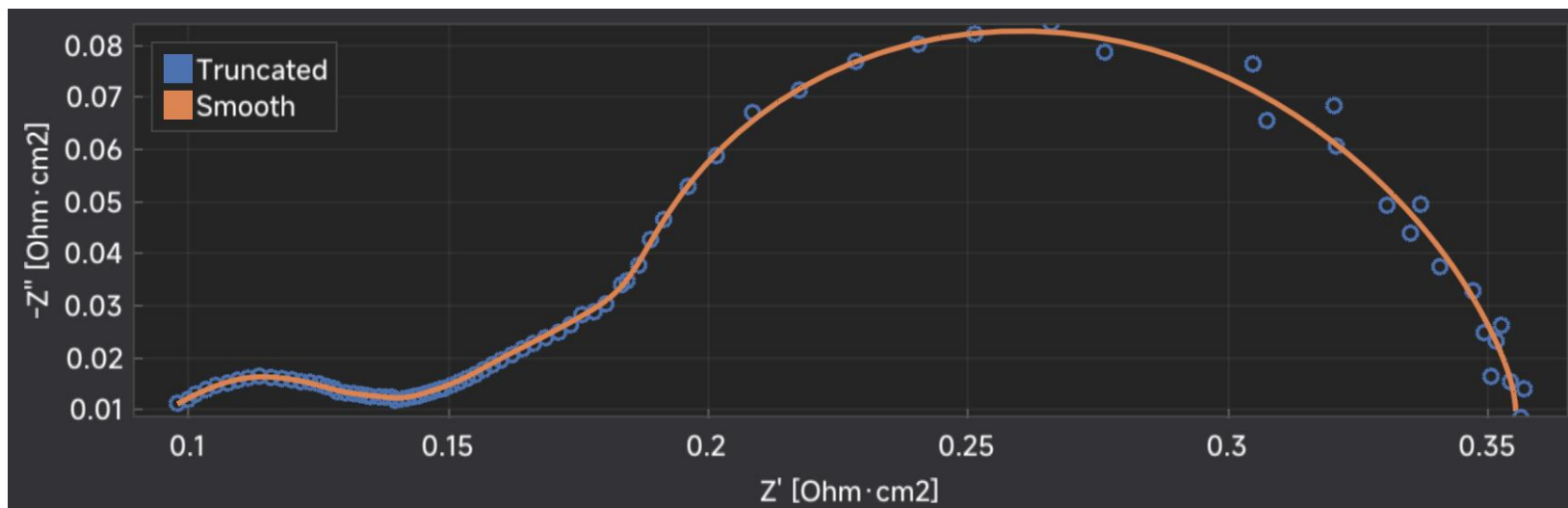
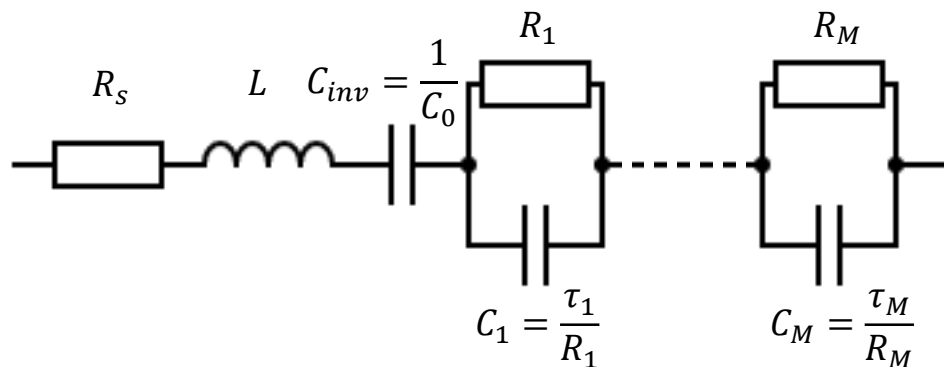
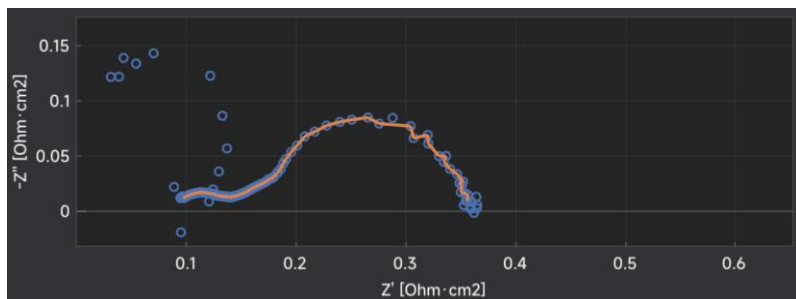
$$\Delta_{Re} = \frac{Z'(\omega) - \hat{Z}'(\omega)}{|Z(\omega)|}$$

$$\Delta_{Im} = \frac{Z''(\omega) - \hat{Z}''(\omega)}{|Z(\omega)|}$$



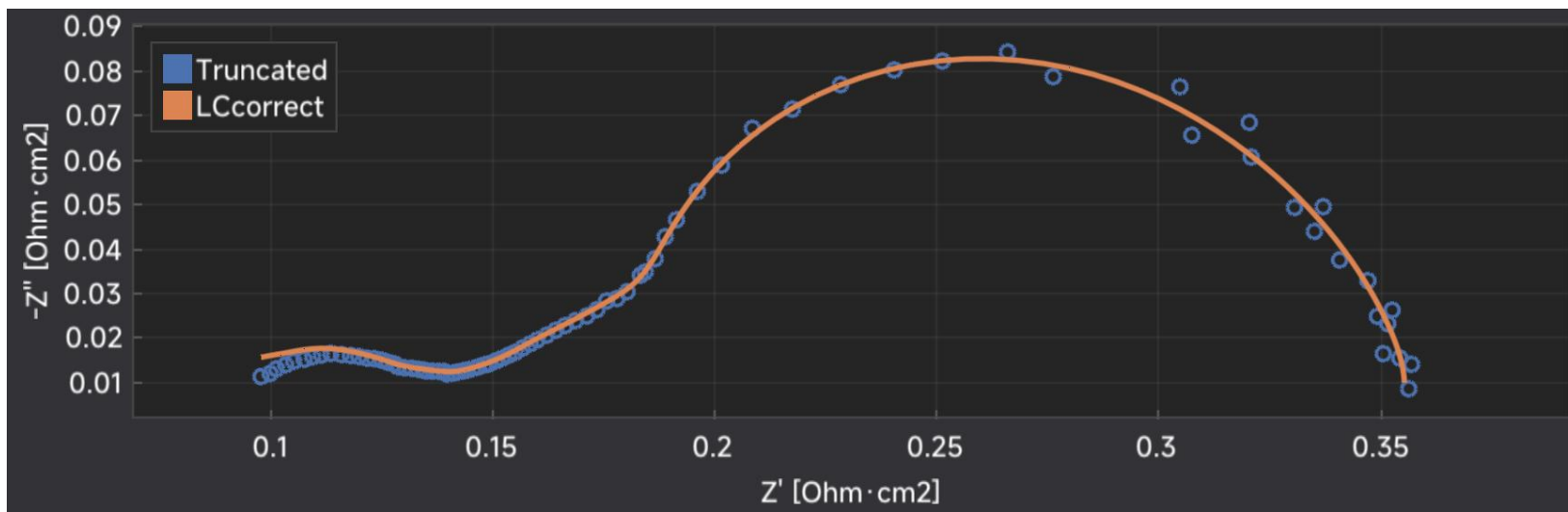
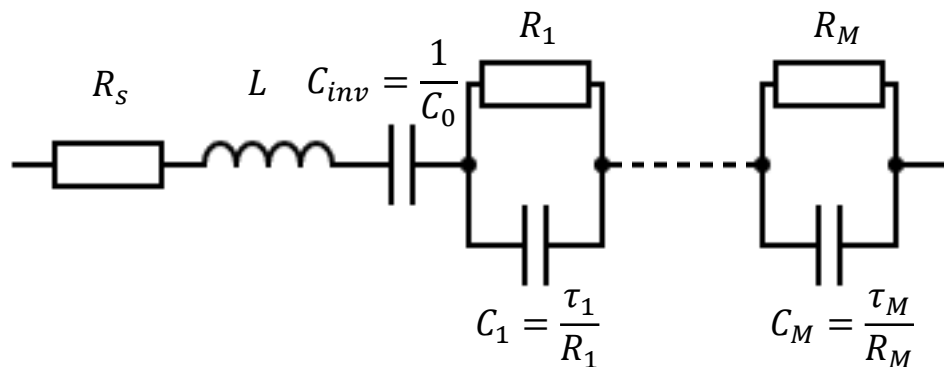
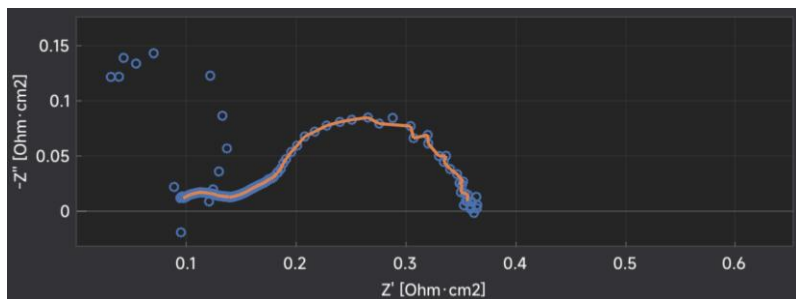
II) Practical aspects, analyses, and applications

Smooth, LC correction, extrapolation based on KK test



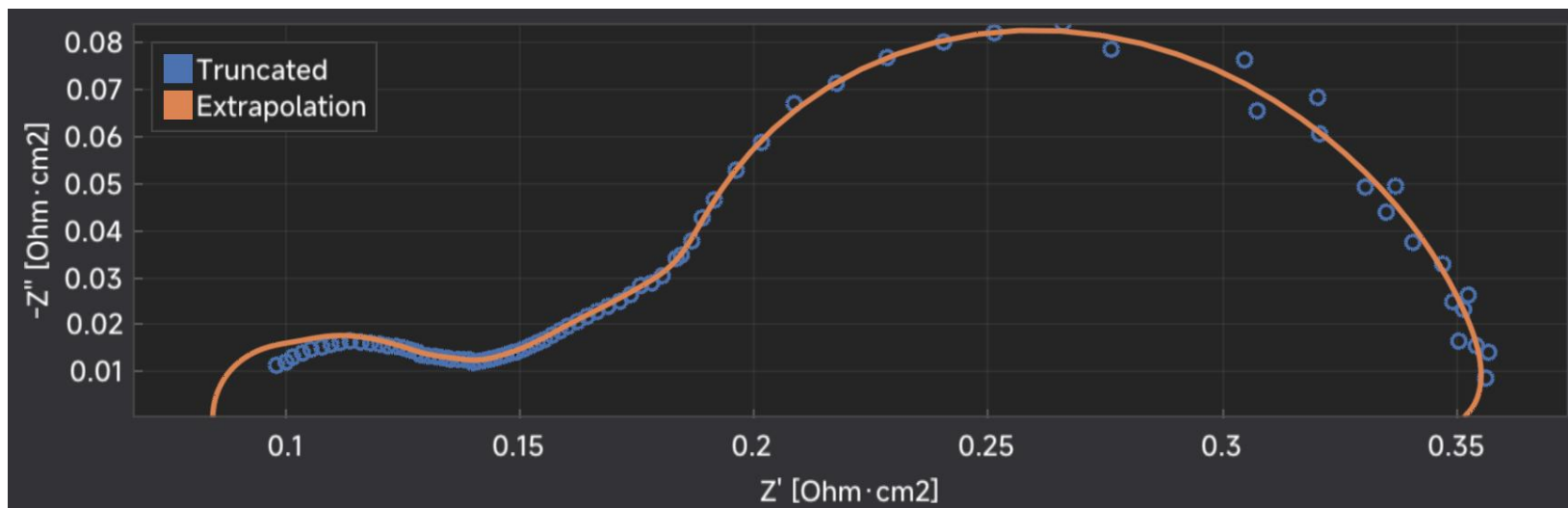
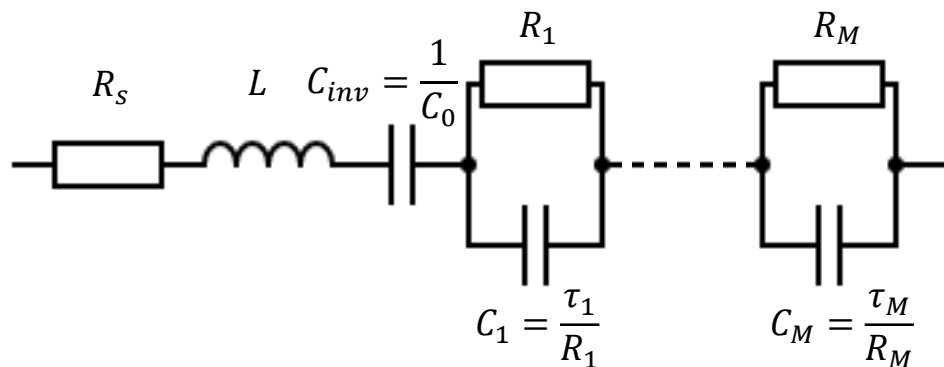
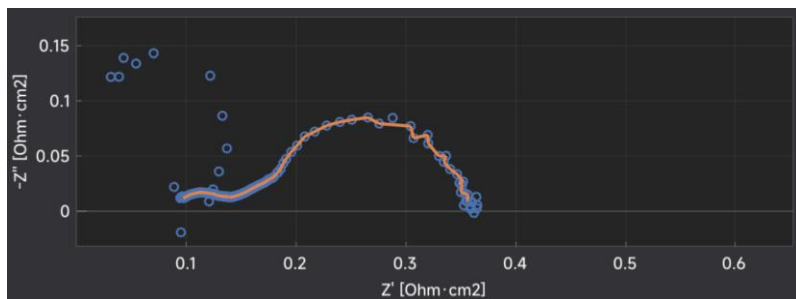
II) Practical aspects, analyses, and applications

Smooth, **LC correction**, extrapolation based on KK test



II) Practical aspects, analyses, and applications

Smooth, LC correction, **extrapolation** based on KK test

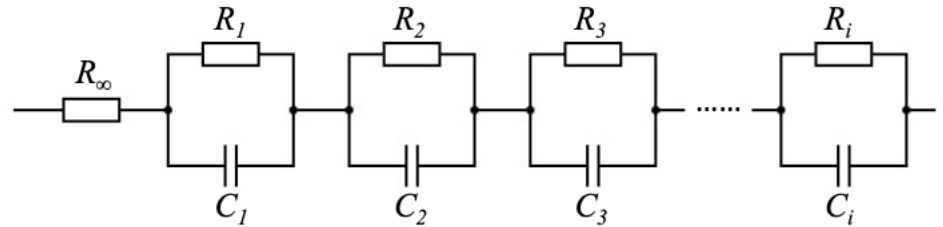


II) Practical aspects, analyses, and applications

Distribution of Relaxation Times (DRT)

$$Z(\omega) = R_{ohm} + \sum_{i=1}^N \frac{R_i}{1 + j\omega\tau_i}$$

$$\tau_i = R_i C_i$$



Generalized Voigt model

Discrete function

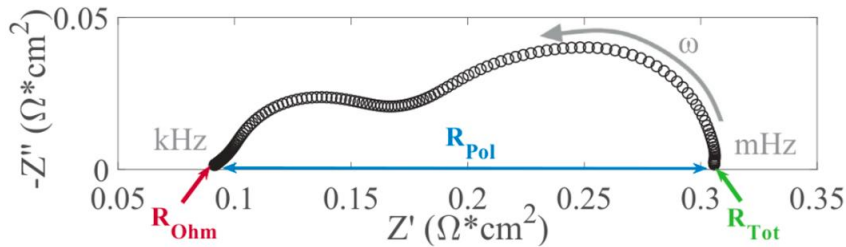
Continuous function

$$Z(\omega) = R_{ohm} + \int_0^{\infty} \frac{R(\tau)}{1 + j\omega\tau} d\tau$$

$$Z(\omega) = R_{ohm} + R_{pol} \int_0^{\infty} \frac{\gamma(\tau)}{1 + j\omega\tau} d\tau$$

$$R(\tau) = R_{pol} \cdot \gamma(\tau)$$

$$\int_0^{\infty} \gamma(\tau) d\tau = 1$$



II) Practical aspects, analyses, and applications

Distribution of Relaxation Times (DRT)

$$Z(\omega) = R_{ohm} + R_{pol} \int_0^{\infty} \frac{\gamma(\tau)}{1 + j\omega\tau} d\tau$$

$$R(\tau) = R_{pol} \cdot \gamma(\tau)$$

$$\int_0^{\infty} \gamma(\tau) d\tau = 1$$

$$Z(\omega) - R_{ohm} = R_{pol} \int_0^{\infty} \frac{\gamma(\tau)}{1 + j\omega\tau} d\tau$$

$$Z(\omega) = Z'(\omega) + jZ''(\omega)$$

$$Z'(\omega) - R_{ohm} = R_{pol} \int_0^{\infty} \frac{\gamma(\tau)}{1 + (\omega\tau)^2} d\tau$$

$$Z''(\omega) = R_{pol} \int_0^{\infty} \frac{\omega\tau \cdot \gamma(\tau)}{1 + (\omega\tau)^2} d\tau$$

Inverse ill-posed problem

- A solution exists;
- The solution is unique
- The solution's behavior changes continuously with the initial conditions

$$Z = A\gamma \quad A = \frac{1}{1 + j\omega\tau}$$

$$\gamma = \arg \min_{\gamma} (\|Z - A\gamma\|^2)$$

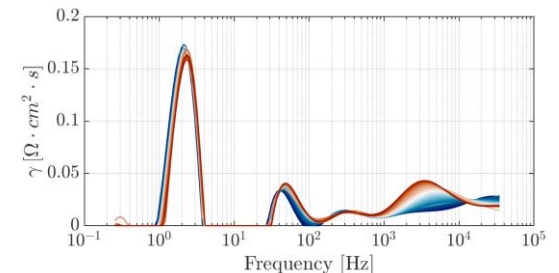
$$\gamma = (A^T A)^{-1} A^T Z$$

Tikhonov regularization (λ as regularization factor)

$$\gamma = \arg \min_{\gamma} (\|A\gamma - Z\|^2 + \lambda \|L\gamma\|^2)$$

$$L = \begin{bmatrix} 1 & 1 & 1 & 1 & \dots \\ 1 & 1 & 1 & 1 & \dots \\ \vdots & \ddots & \ddots & \ddots & \ddots \end{bmatrix}$$

$$\gamma = (A^T A + \lambda L^T L)^{-1} A^T Z$$



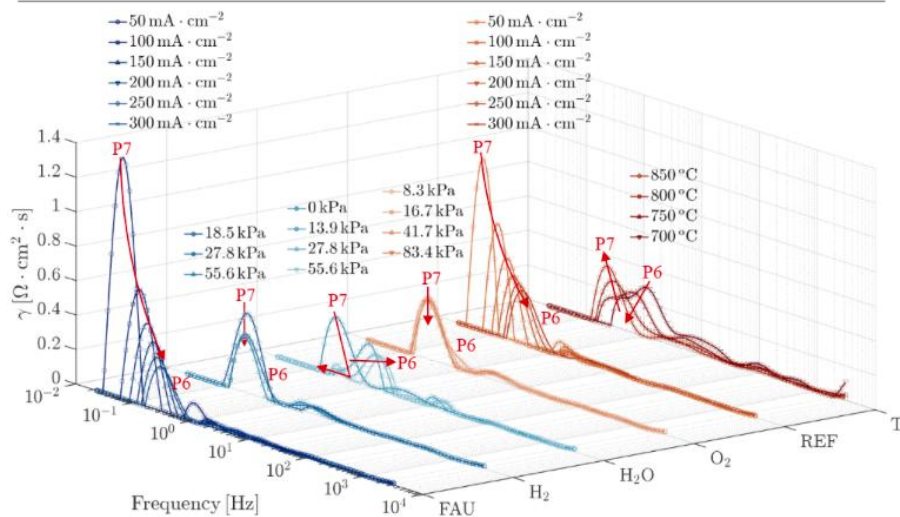
II) Practical aspects, analyses, and applications

DRT for Solid Oxide Cells (650-850 °C)

Cell type 1

	RQ7	RQ6	RQ5	RQ4	RQ3
Gas conversion		O^{2-} surface exchange in GDC	Negatrode and Positrode diffusion	and gas Positrode charge transfer	Negatrode charge transfer
Parameters - statistic	$i > FU > p_{H_2O}$	$T > p_{H_2O}$	$T > p_{H_2} > p_{H_2O}$	$T > AU$	$i > FU > T$
Temperature \uparrow	\swarrow	\checkmark	\checkmark	\downarrow	\swarrow
Current density \uparrow at fixed FU/AU	\swarrow	\swarrow	\swarrow	\swarrow	\swarrow
Current density \uparrow at fixed gas flow	\swarrow	\swarrow	\swarrow	-	-
H₂ partial pressure \uparrow	\downarrow	\rightarrow	-	-	\checkmark
H₂O partial pressure (SH) \uparrow	\swarrow	\swarrow	\swarrow	\leftrightarrow	\uparrow
O₂ partial pressure \uparrow	-	\swarrow	\swarrow	\swarrow	\rightarrow

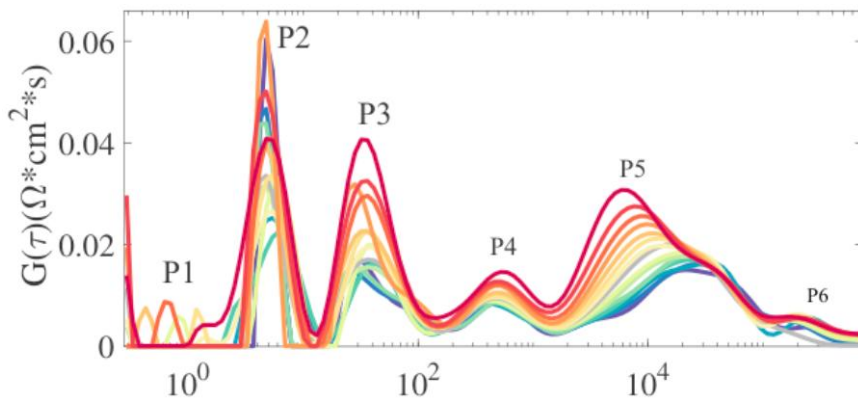
Frequency	0.1-1 Hz	0.1-10 Hz	1-50 Hz (Partially overlapped with O ²⁻ surface exchange)	50-150 Hz	> 150 Hz
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Cell type 2

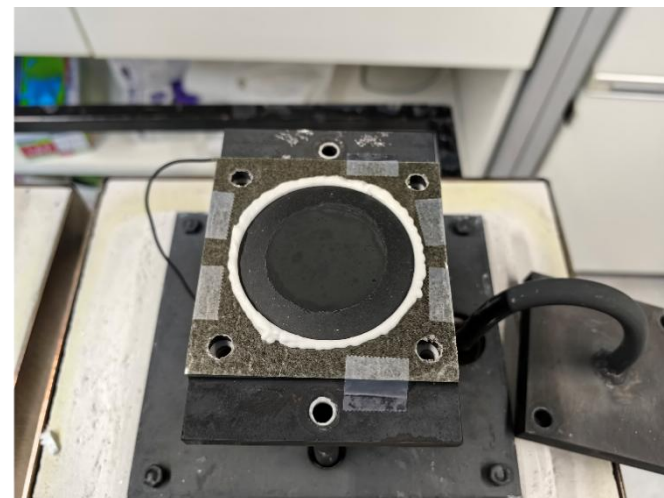
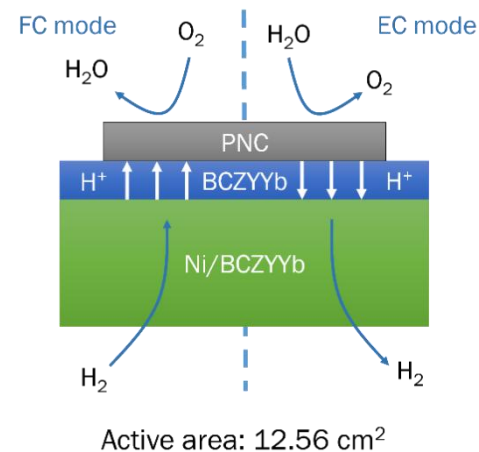
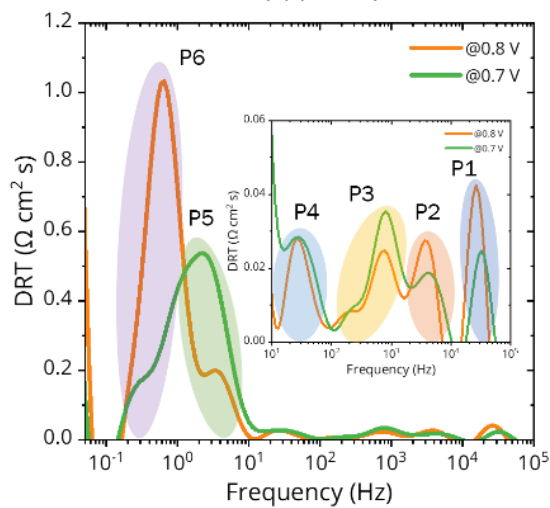
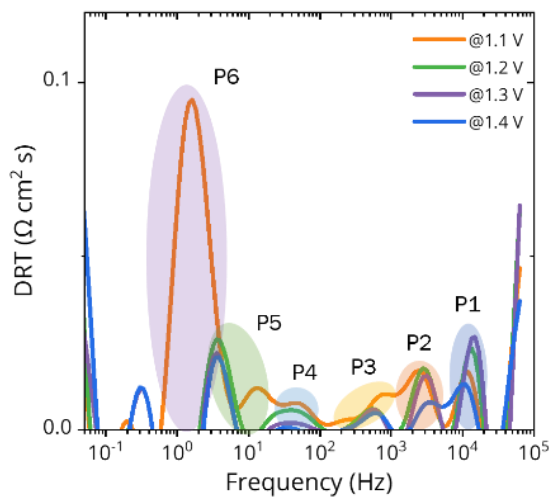
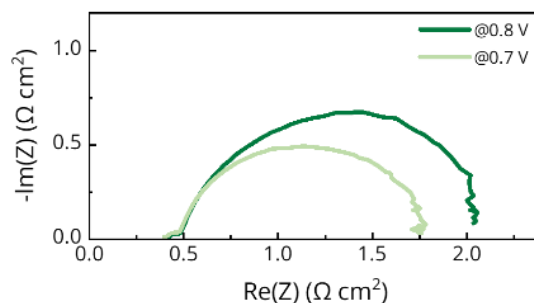
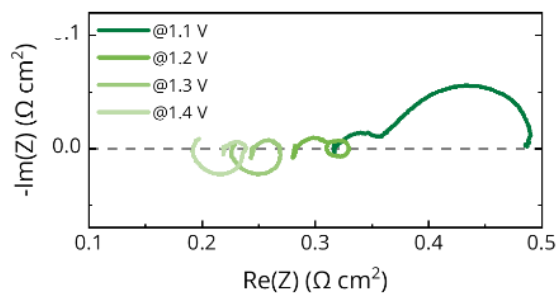
	P1	P2	P3	P4	P5	P6
Conversion diffusion at low pO₂ transport in reforming mixture		Gas conversion	Fuel electrode diffusion/ oxygen electrode reaction + solid state diffusion	Secondary peaks (fuel and oxygen electrode transport)	Fuel electrode charge transfer	High f. peak (?)
Parameters	$p_{O_2} > P$	$F_{tot} > p_{H_2O} > P > i$	$p_{H_2O} > i > p_{O_2} > P$	$p_{H_2O} > T > P$	$T > i > p_{H_2O} > P$	T
Temperature (T) \uparrow	-	-	-	\downarrow	\swarrow	\downarrow
Current density (i) \uparrow	-	-	-	\downarrow	\swarrow	-
Oxygen partial pressure (pO₂) \uparrow	\swarrow	-	\downarrow	\downarrow	\swarrow	-
Steam partial pressure (pH₂O) \uparrow	-	\swarrow	-	\downarrow	\swarrow	-
Total flow rate (F_{tot}) \uparrow	-	\swarrow	-	\downarrow	\swarrow	-
Pressure (P) \uparrow	-	-	\swarrow	\downarrow	\swarrow	-
Frequency	< 1Hz	1-10 Hz	10-100 Hz	0.1-0.5 kHz	0.5-up to 100 kHz *	5- up to 200 kHz *
Single cell no L	0.1	4	50	0.5	20 (2)	200 (6)
Segmented SRU	0.3	4	40	0.3	1	5
Short Stack	0.2	1	20	0.1	0.4	2

* Characteristic frequency influenced by inductance and frequency range
 () values obtained in a different experiment with respect to the sensitivity one



II) Practical aspects, analyses, and applications

DRT for Proton Conducting Cells (400-600 °C)



II) Practical aspects, analyses, and applications

DRT for Proton exchange membrane cells (<100 °C)

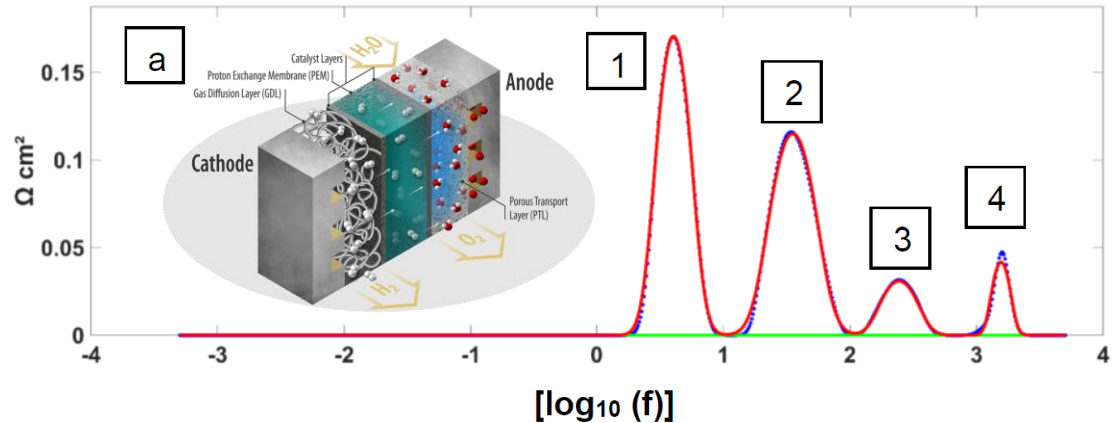
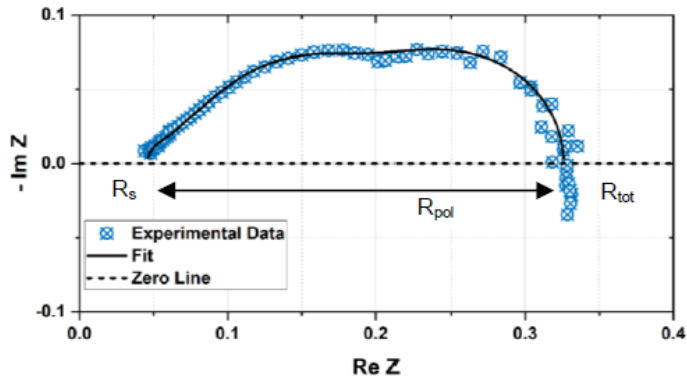
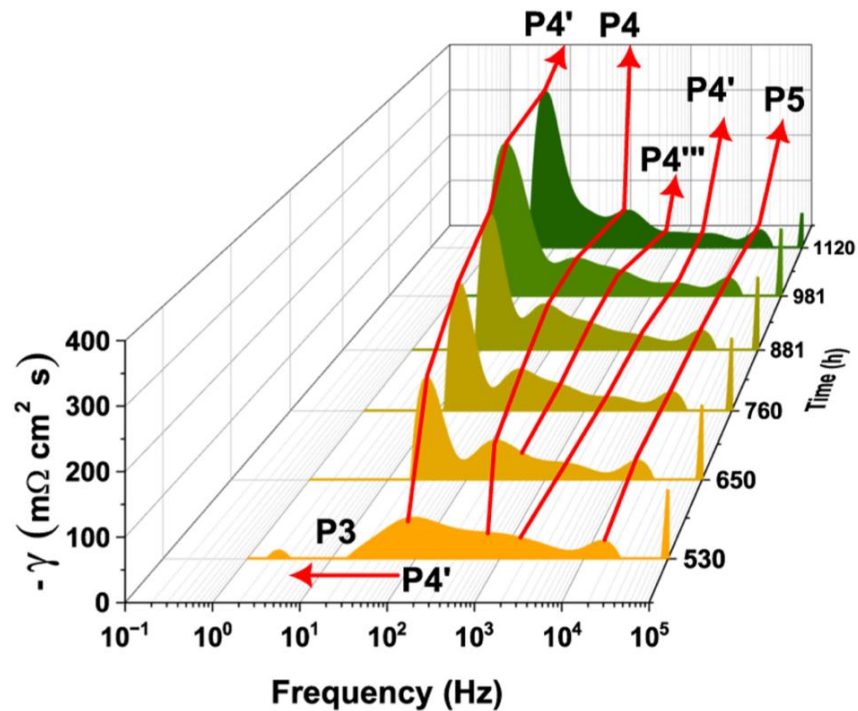
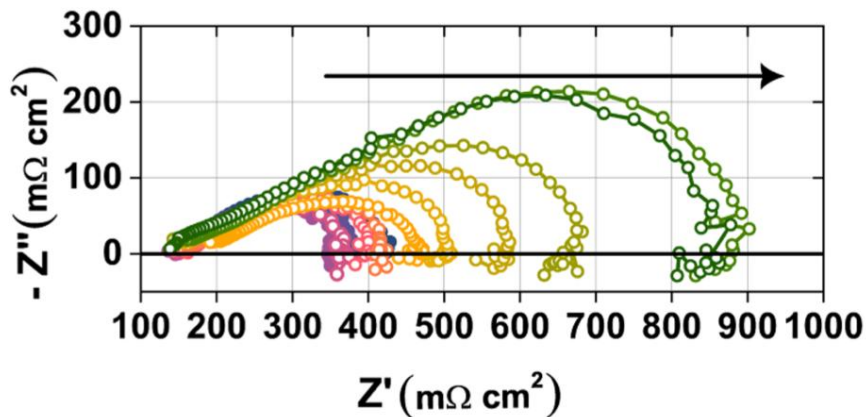


Table 1: The DRT peak positions and the corresponding area in $\Omega \text{ cm}^2$. The peak ASR values are color coded from red (highest) to green (lowest).

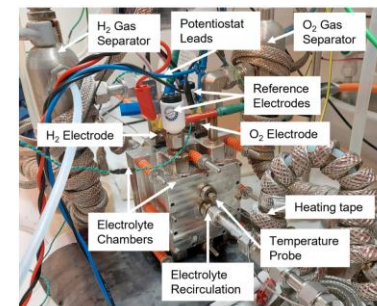
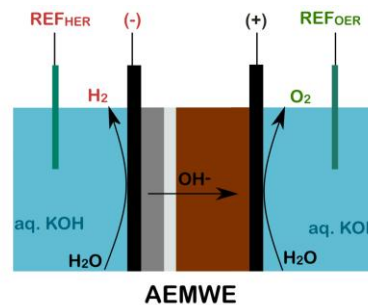
Process Number	Peak Position [$\log_{10} (f)$]	Peak Position [Hz]	Peak Area [$\Omega \text{ cm}^2$]	Literature process assignment	Refs.
1	0.6069	4.0585	0.1282	unknown	[7–10]
2	1.5484	34.1661	0.1142	Gas diffusion	
3	2.3875	242.1380	0.0243	ORR	
4	3.1929	1574.5163	0.0166	H ⁺ transfer	

II) Practical aspects, analyses, and applications

DRT for Anion Exchange Membrane Cells (<100 °C)



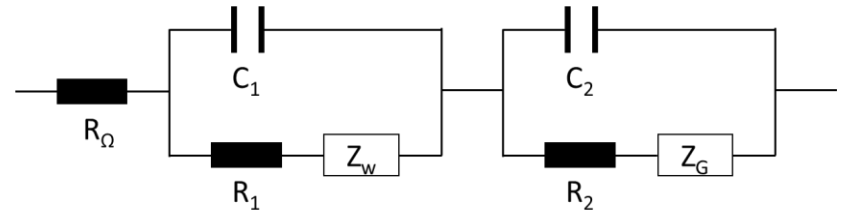
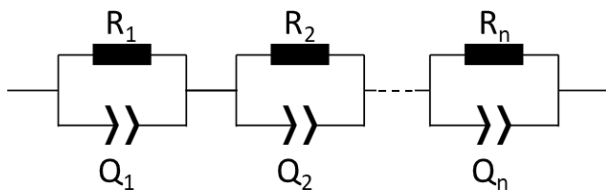
MEA peaks	H ₂ half-cell peaks	O ₂ half-cell peaks	frequency range (Hz)	process description	refs
P1			1–5	water and gas	this work
P2			5–20	diffusion-related	extending ²⁴
P3			25–80	process of HER, OER electrodes	
P4	P4(i) H ₂	P4(i) O ₂ , P4(ii) O ₂	60–2000	HER, OER charge transfer	24
P5		PSO ₂	2200–3200	OER ion transport	this work extending ²⁴
P5	PSH ₂		7000–9000	HER ion transport	this work extending ²⁴



II) Practical aspects, analyses, and applications

2) Data treatment

Complex Nonlinear Least Square fitting (CNLS)



Minimization of the sum of squares function:

$$S = \sum_{i=1}^N w_i \left([Z'_i(\omega_i) - Z'_{M,i}(\omega_i, x)]^2 + [Z''_i(\omega_i) - Z''_{M,i}(\omega_i, x)]^2 \right)$$

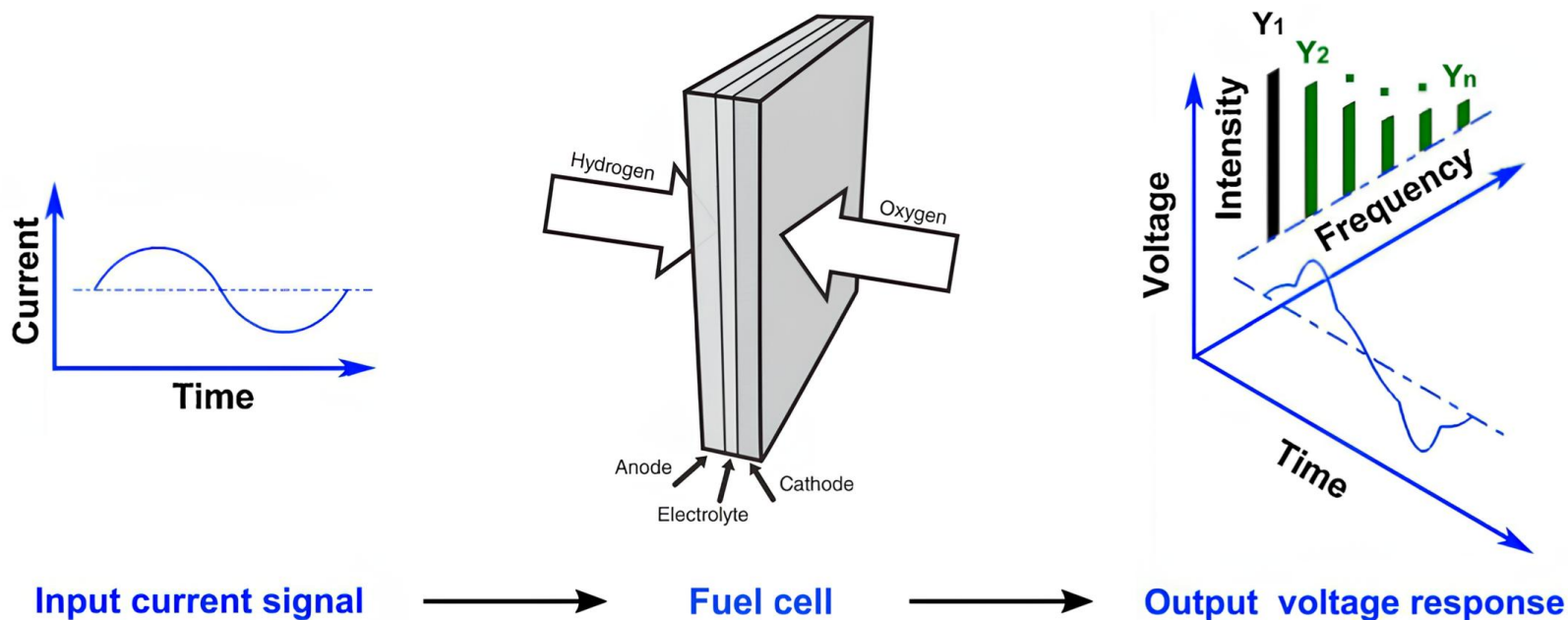
where $Z_i(\omega)$ and $Z_{M,i}(\omega, x)$ are the measured and modeled impedance, respectively

and w_i is the weight factor $w_i = [Z_{Re,i}^2 + Z_{Im,i}^2]^{-1}$

element	impedance	process
(R) serial resistor	R	contact and electrolyte resistance
(RQ) resistor and a constant phase element in parallel	$\frac{R}{1+RQ_0(j\omega)^n}$	double layer capacitance and charge-transfer reactions in parallel
(G) Gerischer element	$\frac{R_{\text{chem}}}{\sqrt{1+j\omega t_{\text{chem}}}}$	combination of electrochemical charge-transfer and diffusion
(W) Warburg element	$R_w \frac{\tanh[(j\omega T)^\alpha]}{(j\omega T)^\alpha}$	diffusion processes
(TLM) transition line model	$\frac{R_{\text{el}}R_{\text{ion,L}}}{R_{\text{el}}+R_{\text{ion,L}}} \left(L + \frac{2\lambda}{\sinh(L/\lambda)} \right) + \lambda_{TLM} \frac{R_{\text{el}}^2+R_{\text{ion,L}}^2}{R_{\text{el}}+R_{\text{ion,L}}} \coth \left(\frac{L}{\lambda_{TLM}} \right)$	porous electrode electrochemistry

II) Practical aspects, analyses, and applications

3) Non linear systems: Total Harmonic Distorsion



$$\text{THD} = \frac{\sqrt{\sum_{n=2}^{\infty} Y_n^2}}{Y_1}$$

